INSTRUCTION MANUAL MODEL 604 DIFFERENTIAL ELECTROMETER AMPLIFIER

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## ILLUSTRATIONS

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## SPECIFICATIONS

-	e ended or differenti				
	o 10r⁴ in 1x and 3	•			
		±1% at dc exclusive of			
		0.1% per 6 months ex			
	LITY: Within 4 mil	livolts per week and 30	00 microvolts/°	C referred to input af	ter a 1-hour
warm-up.		-6.8.10		مقادرت المراجع	
REQUENCY ±3 dB to Maximum lin HIGH FREQ1	RESPONSE: ±1% 200 kHz at mini tear output swing de JENCY ATTENUAT	of full output at do exo do to 10 kHz on all ( imum gain for small s coreases above 10 kHz FOR: 6 dB per octave gives rated frequency m	gains, from 100 signals, decreas to 10% of full with -3 dB po	00-ohm source resista ing to 50 kHz on ma i output at 75 kHz.	ximum gain.
NOISE: Less	than 20 microvolts	rms, 3 to 100 Hz; les	s than 60 mic	ovolts rms, 3 Hz to i	50 kHz with
maximum g	ain (approximately	triple with minimum g			
	decrease when guar				
		an 10 <sup>14</sup> ohms shunted			
	en in guarded positi	ed by leas than 1 picc	otarad in guarde	a position. Inner shie	IC OF TRAXIE
		Greater than 90 dB, dc	to 120 Hz de	creasing to SO dB at	1 kHz
		Ip to $\pm 11$ volts, dc t			
at 75 kHz.	COLINCE. U			The real ways	
	O volte single ende	d for full scale at up to	5 milliamperer	B.	
Unity Gain	: Two outputs pro	ovide signals equal (	to each input-t		005% at do
(exclusive o	f up to 5 millivolts z	ero offset referred to d	output).	-	
All outputs	are short-circuit proc	<b>л.</b>			
<u>as an am</u> n	ETER (with access	lory shunts):			
	Less than 2	x 10 <sup>-14</sup> ampere.			
		SE (with Model 6041	Differential Cur	rent Shunt*):	
YPICAL RIS	E TIMES AND NO				
hunt Resistor	Current Span	Gu	rded	Ung	uarded
Shunt Realstor (ohms)	Current Span (full scale)	Gus Rise Time (10%-90%)	Noise (p-p)	Ung Rise Time (10%-90%)	Noise (p-p
Shunt Resistor (ohms) 10*	Current Span (full scale) 10** to 10** A	Gus Rise Time (10%-90%) 0.00005 a	Noise (p-p) 2x10* A	Ung Rise Time (10%-90%) 0.0005 •	Noise (p-p 2x10** A
Shunt Resistor (ohms) 10* 10* 10*	Current Span (full scale) 10 <sup>-e</sup> to 10 <sup>-e</sup> A 10 <sup>-e</sup> to 10 <sup>-e</sup> A 10 <sup>-te</sup> to 10 <sup>-ta</sup> A	Gua Rise Time (10%-90%) 0.00005 s 0.0005 s 0.05 s	Noise (p-p) 2x10* A 2x10* A 2x10** A 2x10** A	Ung Rise Time (10%-90%) 0.0005 e 0.05 e 5.0 s	Noise (p-p 2x10 <sup>-10</sup> A 2x10 <sup>-10</sup> A 2x10 <sup>-14</sup> A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 10* 10* 10* 10*	Current Span (full scale) 10* to 10* A 10* to 10* A	Gus Rise Time (10%-90%) 0.00005 s 0.0005 s	Noise (p-p) 2x10 <sup>-9</sup> A 2x10 <sup>-19</sup> A 2x10 <sup>-19</sup> A 2x10 <sup>-13</sup> A 2x10 <sup>-13</sup> A	Ung Rise Time (10%-90%) 0.0005 e 0.05 e 5.0 s 50.0 s suring current with the Inte	Noise (p-p 2x10-** A 2x10-** A 2x10-** A 2x10-** A 2x10-** A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 604 Model 604	Current Span (full scale) 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A et of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: uracy: = 2% 10* to 10	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.5 s and no other shunt capac b4 has slightly shorter rise ant Shunt " ohms in decade steps.	Noles (p-p) 2x10* A 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-15</sup> A 2x10 <sup>-15</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 604 Model 604	Current Span (full scale) 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A et of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: uracy: = 2% 10* to 10	Gus         Gus           Rise Time (10%-90%)         0.0005 s           0.00005 s         0.0005 s           0.05 s         0.05 s           0.5 s         s           and no other shunt capace         04 has slightly shorter rise           ant Shunt         0.0000	Noles (p-p) 2x10* A 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-15</sup> A 2x10 <sup>-15</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohme) 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 604 Model 604 Model 604 Connectors: Dimensione, 1	Current Span (full scale) 10° to 10° A 10° to 10° A et of triaxisi input cable Resistore, the Model 60 1 Differential Curre Specifications: Jracy: ± 2% 10° to 10 Input: Triaxisi. Output: Weight: 7° high x 8%	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.5 s and no other shunt capac b4 has slightly shorter rise ant Shunt " ohms in decade steps.	Noles (p-p) 2x10* A 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohms) 10* 10* 10'* 10'* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Acta Connectors: Dimensions, ' S A DC VO	Current Span (full scale) 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A st of triaxisl input cable Resistors, the Model 60 1 Differential Curre Bpecifications: Jracy: ± 2% 10* to 10 Input; Triaxisl. Output: Weight: 7" high x 8% LTMETER:	Gua Rise Time (10%-90%) 0.0005 s 0.005 s 0.5 s and no other shunt capac bas alightly shorter rise ant Shunt I'' ohms in decade steps. Cable and connectors mate '' wide x 10'' deep; net w	Noles (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup>	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* * with three fe 6033 Shunt ( ranges. Model 604 Model 6041 Resistor Acci Connectors: Dimensions, f S A DC VO ANGE: ±1 T	Current Span (full scale) 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A et of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: Jracy: ±2% 10 <sup>4</sup> to 10 Input: Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to	Gus           Rise Time (10%-90%)           0.00005 s           0.005 s           0.05 s           0.5 s           9 and no other shunt capacity           94 has slightly shorter rise           95 of the short short           96 of the short short           97 of the short short           1 volt in seven 1x s	Noise (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A citance. When mea times and lower with Model 804 reight, 6 pounds. and 3x steps.	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-10</sup> A 2x10 <sup>-10</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Accu Connectors: Dimensions, ' S A DC VO ANGE: ±1 m IETER ACCU	Current Span (full scale) 10 <sup>-6</sup> to 10 <sup>-10</sup> A 10 <sup>-10</sup> to 10 <sup>-11</sup> A 10 <sup>-10</sup> to 10 <sup>-11</sup> A 10 <sup>-11</sup> to 10 <sup>-11</sup> A et of triaxisl input cable Resistors, the Model 60 1 Differential Curre Specifications: Jracy: ± 2% 10 <sup>4</sup> to 10 Input: Triaxisl, Output: Weight: 7 <sup>11</sup> high x 8 <sup>3</sup> / <sub>4</sub> LTMETER: nillivolt full scale to RACY: ± 2% of ful	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.05 s 0.05 s and no other shunt capac b4 has alightly shorter rise ant Shunt " ohns in decade steps. Cable and connectors mate " wide x 10" deep; net w ± 1 volt in seven 1x a Il scale exclusive of noi	Noise (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A citance. When mea times and lower with Model 804 reight, 6 pounds. and 3x steps.	Ung Rise Time (10%-90%) 0.0006 e 0.05 e 5.0 e 50.0 e suring current with the Inter noise for corresponding	Noise (p-p 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A 2x10 <sup>-16</sup> A
Shunt Resistor (ohms) 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Accu Cannectors: Dimensions, S A DC VO ANGE: ±1 m IETER ACCU IETER NOISE	Current Span (full scale) $10^{-6}$ to $10^{-11}$ A $10^{-6}$ to $10^{-11}$ A $10^{-10}$ to $10^{-12}$ A $10^{-11}$ to $10^{-12}$ A et of triaxial input cable Resistors, the Model 60 <b>1 Differential Curre</b> Specifications: uracy: $\pm 2\%$ 10 <sup>4</sup> to 10 input: Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full $\pm \pm 40$ microvolts v	Gua Rise Time (10%-90%) 0.00005 s 0.0005 s 0.05 s	Noles (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-13</sup> A 2x10 <sup>-13</sup> A citance. When mea times and lower with Model 604 reight, 6 pounds. and 3x steps. se and drift.	Ung Rise Time (10%-90%) 0.0005 = 5.0 s 50.0 s suring current with the inte- noise for corresponding input.	Noise (p.g. 2x10 <sup>-13</sup> A 2x10 <sup>-13</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-16</sup> A moley mounted Model 8041
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 6041 Resistor Acc Connectors: Dimensions OI ANGE: ±10 IETER ACCU IETER NOISE OLARITY: M	Current Span (full scale) 10 <sup>-6</sup> to 10 <sup>-6</sup> A 10 <sup>-8</sup> to 10 <sup>-11</sup> A 10 <sup>-10</sup> to 10 <sup>-11</sup> A 10 <sup>-11</sup> to 10 <sup>-14</sup> A et of triaxisl input cable Resistore, the Model 60 1 Differential Curre Bpecifications: uracy: ±2% 10 <sup>4</sup> to 10 input; Triaxisl. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: ±2% of full : ±40 microvolts v leter switch select	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.05 s 0.05 s 0.05 s and no other shunt capac 4 has slightly shorter rise ent Shunt <sup>11</sup> ohms in decade steps. Cable and connectors mate " wide x 10" deep; net w ± 1 volt in seven 1x s it is cale exclusive of noi vith inputs shorted. s left-zero (positive of	Noles (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-13</sup> A 2x10 <sup>-13</sup> A citance. When mea times and lower with Model 604 reight, 6 pounds. and 3x steps. se and drift.	Ung Rise Time (10%-90%) 0.0005 = 5.0 s 50.0 s suring current with the inte- noise for corresponding input.	Noise (p-p 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Model 6041
Shunt Resistor (ohme) 10* 10* 10* 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 604 Resistor Acc Connectors: Dimensions. S A DC VO ANGE: ±1 m IETER ACCU IETER ACCU IETER NOISE OLARITY: M does not reve	Current Span (full scale) $10^{-6}$ to $10^{-11}$ A $10^{-6}$ to $10^{-11}$ A $10^{-10}$ to $10^{-12}$ A $10^{-11}$ to $10^{-12}$ A et of triaxial input cable Resistors, the Model 60 <b>1 Differential Curre</b> Specifications: uracy: $\pm 2\%$ 10 <sup>4</sup> to 10 input: Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full $\pm \pm 40$ microvolts v	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.05 s 0.05 s 0.05 s and no other shunt capac 4 has slightly shorter rise ent Shunt <sup>11</sup> ohms in decade steps. Cable and connectors mate " wide x 10" deep; net w ± 1 volt in seven 1x s it is cale exclusive of noi vith inputs shorted. s left-zero (positive of	Noles (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-13</sup> A 2x10 <sup>-13</sup> A citance. When mea times and lower with Model 604 reight, 6 pounds. and 3x steps. se and drift.	Ung Rise Time (10%-90%) 0.0005 = 5.0 s 50.0 s suring current with the inte- noise for corresponding input.	Noise (p-p 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Model 6041
Shunt Resistor (ohme) 10* 10* 10* 10* 10* 10* 6033 Shunt I ranges. Model 604 Resistor Acc Connectors: Dimensions. S A DC VO ANGE: ±1m IETER ACCU IETER NOISE OLARITY: M does not revi ENERAL:	Current Span (full scale) 10-4 to 10-4 A 10-4 to 10-4 A 10-4 to 10-4 A 10-4 to 10-4 A 10-4 to 10-4 A et of triaxisl input cable Resistore, the Model 60 1 Differential Curre Bpecifications: aracy: ±2% 104 to 10 input: Triaxisl. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: ±2% of ful :: ±40 microvolts v leter switch select arse polarity of output	Gua Rise Time (10%-90%) 0.00005 s 0.005 s 0.05 s 0.5 s and no other shunt capac 4 has slightly shorter rise and Shunt other shunt capac the shunt shore shunt capac and shunt other shunt and shunt capac and shunt and and shunt and and and and and and and and and and	Noles (p-p) 2x10 <sup>-1</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-11</sup> A 2x10 <sup>-13</sup> A Ditance. When mea times and lower with Model 604 (eight, 6 pounds. and 3x steps. as and drift. or negative) or	Ung Rise Time (10%-90%) 0.0005 = 5.0 s 50.0 s suring current with the inte- noise for corresponding input.	Noise (p-p 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Model 6041
Shunt Resistor (ohms) 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Accu Connectors: Dimensions, f S A DC VO ANGE: ±1 m IETER ACCU IETER ACU IETER NOISE OLARITY: M does not revi ENERAL: ERO SUPPRI	Current Span (full scale) 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A 10* to 10* A et of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: uracy: $\pm 2\%$ 10° to 10 input: Triaxial. Output: Weight: 7" high x 8% LTMETER: millivolt full scale to RACY: $\pm 2\%$ of full :: $\pm 40$ microvolts v leter switch select prise polarity of output: ESSION: Up to $\pm 1$	Gus           Rise Time (10%-90%)           0.0006 s           0.005 s           0.05 s           0.5 s           and no other shunt capacity           Pane alightly shorter rise           and no other shunt capacity           Pane alightly shorter rise           and the cade steps.           Cable and connectors mate           "wide x 10" deep; net w           ± 1 volt in seven 1x s           Il scale exclusive of noi           vith inputs shorted.           s left-zero (positive outs.           volt dc between inputs	Noles (p-p) 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A bitance. When mea times and lower with Model 604 reight, 6 pounds. and 3x steps. se and drift. or negative) or	Ung Rise Time (10%-90%) 0.0005 = 5.0 = 50.0 = suring current with the Inte noise for corresponding input.	Noise (p-p 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Model 6041 Model 6041
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 604 Model 6041 Resistor Accu Connectors: Dimensions, S A DC VO ANGE: ±1 m IETER ACCU IETER NOISE OLARITY: M does not revi ENERAL: ERO SUPPRI IAXIMUM IN	Current Span (full scale) 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A et of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: uracy: $\pm 2\%$ 10 <sup>4</sup> to 10 Input: Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full $\vdots \pm 40$ microvolts w leter switch select arse polarity of output: ESSION: Up to $\pm 1$ UPUT OVERLOAD:	Gus           Rise Time (10%-90%)           0.0006 s           0.005 s           0.05 s           0.5 s           and no other shunt capacity           Pand by the shunt capacity	Notes (p-p) 2x10* A 2x10**A 2x10**A 2x10**A 2x10**A 2x10**A Control A Control A	Ung Rise Time (10%-90%) 0.0005 = 0.05 = 5:0 = suring current with the inte- noise for corresponding input. center-zero scales. No t to ground or between	Noise (p-p 2x10 <sup>-13</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-16</sup> A maily mounted Model 6041
Shunt Resistor (ohms) 10* 10* 10* 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Accu Connectors: Dimensions, / S A DC VOI ANGE: ±1 m IETER ACCU IETER NOISE OLARITY: M does not revi ENERAL: ERO SUPPRI IAXIMUM IN ONNECTORS	Current Span (full scale) 10-4 to 10-1 A 10-4 to 10-1 A 10-1 to 10-1 A 10-1 to 10-1 A 10-11 to 10-14 A at of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: araoy: $\pm 2\%$ 10 <sup>4</sup> to 10 Input: Triaxial. Output: Weight: 7'' high x 8 <sup>3</sup> /4 LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full $\vdots \pm 40$ microvolts v leter switch select arase polarity of output ESSION: Up to $\pm 1$ JPUT OVERLOAD: 3: Input: Special T	Gus           Rise Time (10%-90%)           0.0006 s           0.005 s           0.05 s           0.5 s           and no other shunt capacity           Pane alightly shorter rise           and no other shunt capacity           Pane alightly shorter rise           and the cade steps.           Cable and connectors mate           "wide x 10" deep; net w           ± 1 volt in seven 1x s           Il scale exclusive of noi           vith inputs shorted.           s left-zero (positive outs.           volt dc between inputs	Notes (p-p) 2x10* A 2x10**A 2x10**A 2x10**A 2x10**A 2x10**A Control A Control A	Ung Rise Time (10%-90%) 0.0005 = 0.05 = 5:0 = suring current with the inte- noise for corresponding input. center-zero scales. No t to ground or between	Noise (p-p 2x10-19 A 2x10-19 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A Model 6041
Shunt Resistor (ohms) 10* 10* 10" * with three fe 6033 Shunt I ranges. Model 6041 Resistor Aca Connectors: Dimensions, ' S A DC VO ANGE: ±1 m IETER ACCU IETER ACU IETER ACUSE OLARITY: M does not revi ENERAL: ERO SUPPRI IAXIMUM INONNECTORS	Current Span (full scale) 10* to 10* A 10* to 10* A st of triaxial input cable Resistors, the Model 60 1 Differential Curre Bpacifications: uracy: $\pm 2\%$ 10* to 10 input: Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full $\vdots \pm 40$ microvolts w elter switch select arse polarity of output: ESSION: Up to $\pm 1$ JPUT OVERLOAD: $\vdots$ Input: Special T tiput: BNC.	Gua Rise Time (10%-90%) 0.0005 s 0.005 s 0.05 s 0.5 s and no other shunt capac 4 has alightly shorter rise and Shunt 11 ohms in decade steps. Cable and connectors mate 11 scale exclusive of noi wide x 10" deep; net w ± 1 volt in seven 1x a 11 scale exclusive of noi vith inputs shorted. s left-zero (positive o uts. volt dc between inputs ± 400 volts, dc - 100 felon-insulated Triaxia	Noles (p-p) 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Ditance. When mea times and lower with Model 604 with Model 604 wi	Ung Rise Time (10%-90%) 0.0005 = 5.0 = 50.0 = suring current with the inter- noise for corresponding input. center-zero scales. N t to ground or betwee 0-1. Amplifier output:	Noise (p-p 2x10-19 A 2x10-19 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A Model 6041
Shunt Resistor (ohms) 10* 10* 10" * with three fe 6033 Shunt I ranges. Model 604 Model 6041 Resistor Accu Connectors: Dimensions, ' S A DC VO ANGE: ±1 m IETER ACCU IETER ACCU IETER ACUSE OLARITY: M does not revi ENERAL: ERO SUPPRI IAXIMUM IN ONNECTORS Unity-gain ou OWER: 105-	Current Span (full scale) 10* to 10* A 10* to 10* A st of triaxial input cable Resistors, the Model 60 1 Differential Curre Bpacifications: Jracy: $\pm 2\%$ 10* to 10 Input; Triaxial. Output: Weight: 7" high x 8% LTMETER: nillivolt full scale to RACY: $\pm 2\%$ of full E: $\pm 40$ microvolts v leter switch select prise polarity of output ESSION: Up to $\pm 1$ IPUT OVERLOAD: S: Input: Special T tiput: BNC. 125 or 210-250 vol	Gua Rise Time (10%-90%) 0.0005 s 0.005 s 0.05 s 0.5 s and no other shunt capac 0.4 has alightly shorter rise and Shunt other in decade steps. Cable and connectors mate "wide x 10" deep; net w ± 1 volt in seven 1x s 1 scale exclusive of noi vith inputs shorted. s left-zero (positive of uts. volt dc between inputs ±400 volts, dc - 100 effon-insulated Triaxia bits (switch selected); E	Noles (p-p) 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Ditance. When mea times and lower with Model 604 (eight, 6 pounds. and 3x steps. as and drift. or negative) or kHz either inpu a) Dage 33050 50-60 Hz; 10 w	Ung Rise Time (10%-90%) 0.0005 = 0.05 = 5.0 = suring current with the inter- noise for corresponding input. center-zero scales. N t to ground or betwee -1. Amplifier output: ratts.	Noise (p-p 2x10 <sup>-13</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-14</sup> A 2x10 <sup>-16</sup> A maily mounted Model 6041
Bhunt Resistor (ohms) 10* 10* 10* 10* 10* * with three fe 6033 Shunt I ranges. Model 6041 Resistor Accu Connectors: Dimensions, f S A DC VO ANGE: ±1 m IETER ACCU IETER ACCU IETER ACU IETER ACUSE OLARITY: M does not revo ENERAL: ERO SUPPRI IAXIMUM IN ONNECTORS Unity-gein ou OWER: 105- IMENSIONS	Current Span (full scale) 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A 10-* to 10-* A at of triaxial input cable Resistors, the Model 60 1 Differential Curre Specifications: Jracy: $\pm 2\%$ 10* to 10 Input: Triaxial. Output: Weight: 7" high x 8% LTMETER: millivolt full scale to RACY: $\pm 2\%$ of full E: $\pm 40$ microvolts v leter switch select prise polarity of output ESSION: Up to $\pm 1$ IPUT OVERLOAD: S: Input: Special T tiput: BNC. 125 or 210-250 vc, WEIGHT: 7" high	Gua Rise Time (10%-90%) 0.0005 s 0.005 s 0.05 s 0.5 s and no other shunt capac 4 has alightly shorter rise and Shunt 11 ohms in decade steps. Cable and connectors mate 11 scale exclusive of noi wide x 10" deep; net w ± 1 volt in seven 1x a 11 scale exclusive of noi vith inputs shorted. s left-zero (positive o uts. volt dc between inputs ± 400 volts, dc - 100 felon-insulated Triaxia	Noles (p-p) 2x10* A 2x10** A 2x10** A 2x10** A 2x10** A 2x10** A Ditance. When mea times and lower with Model 604 (eight, 6 pounds. and 3x steps. as and drift. or negative) or kHz either inpu bage 33050 50-50 Hz; 10 w	Ung Rise Time (10%-90%) 0.0005 = 5.0 = 50.0 = suring current with the Inter- noise for corresponding input. center-zero scales. N t to ground or betwee -1. Amplifier output: ratts. , 12 pounds.	Noise (p-p 2x10-19 A 2x10-19 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A 2x10-14 A Model 6041

### SECTION 1. GENERAL DESCRIPTION

1-1. GENERAL.

a. The Keithley Model 604 Differential Electrometer Amplifier is a completely solid-state instrument that can be used by itself single-endedly or differentially either as an amplifier or a millivoltmeter. The Amplifier has selectable gains of 10 to  $10^4$  in 1x and 3.33x steps corresponding to the VOLTS Switch settings of 1 to .001.

b. When used in conjunction with the Model 6041 Differential Current Shunt the Model 604 becomes a single-ended or differential picoammeter. Single-ended and differential current measurements from  $10^{-6}$  to  $10^{-14}$  ampere can be made with this setup. Also, high megohm resistors may be installed within the Model 604 to allow shunt picoammeter operation.

1-2. MODEL 604 FEATURES.

a. As an amplifier, the Model 604 will operate over a bandwidth from dc to 50 kHz or greater with six selectable 3-dB points from 30 kHz to 100 Hz. Thus, the signal-to-noise ratio may be optimized for a particular application. The amplifier output furnishes ±10 volts at 5 milliamperes single-ended.

b. The Electrometer Amplifier has an input resistance of greater than  $10^{14}$  ohms shunted by less than 5 picofarads in the unguarded position, that allows high impedance measurement to be made easily. When in the guarded position the input impedance is shunted by less than 1 picofarad.

c. Low offset current - less than  $2 \times 10^{-14}$  ampere - minimizes zero offset with high source resistance and permits maximum resolution when measuring current.

d. A choice of outputs which can be used simultaneously or singly is available. Each input of the Model 604 has a unity gain output, which has ±50 ppm accuracy. This allows the Model 604 to be used as a preamplifier with differential or digital voltmeters for precise measurements. Also, these outputs can be fed into an X-Y recorder for recording data from high impedance sources. When desired, the unity-gain outputs can be monitored for absolute values while recordings are made from the amplifier output.

e. All outputs are short-circuit proof, and output noise referred to the input is nearly constant regardless of the gain.

f. The guarding capability of the Model 604 allows fast measurements from high resistance sources, by eliminating the effects of cable capacitance and leakage. To conveniently facilitate these measurements, the Model 6301 Guarded Probe can be used to connect the source to the Model 604 in the guarded mode. The guard feature is switch selectable. g. Another outstanding feature of the Keithley Model 604 is its zero stability. Signals can be monitored over weeks without constant rezeroing. Zero drift of the Amplifier is less than 4 millivolts per week; zero offset due to temperature changes is less than 300 microvolts per  $^{\rm O}$ C. Zero shift due to mechanical shock or vibration is negligible.

h. Overloads up to ±400 volts will not damage the 604 Amplifier, and recovery is almost immediate. A unique circuit provides protection while maintaining the favorable features of the MOS FET input.

i. As another convenience feature, the Model 604 provides a very stable ±1 volt suppression on any range. Variations as small as 1 millivolt in a 1-volt signal can be displayed full scale.

j. Variations in line voltages from 105 to 125 volts cause no apparent change even on the most sensitive range. Line frequency noise is practically eliminated from the Model 604 output.

1-3. MODEL 6041. The Model 6041 Differential Curent Shunt is an accessory specifically designed to convert the Model 604 Amplifier into a single-ended or differential multi-range picoammeter.

a. The driven guard enables the Model 604 Amplifier and the Model 604l Shunt combination to obtain fast response in current measurements. Input capacitance at the end of a 10-foot cable is maintained at 1 picofarad in the guarded mode. As a result, rise times of 0.5 millisecond are possible with  $10^{-11}$  ampere input signals. Guarding is maintained through the Model 604l by the use of triaxial input connectors and total guarding within the Shunt itself.

b. For those applications where maximum signalto-noise ratio is more desirable than fast response, the Model 604 can be used in its unguarded mode. Noise will be as low as  $2 \times 10^{-15}$  ampere using a  $10^{-1}$  ohm resistor. However, rise times are much longer than when the Model 604 is in its guarded mode. Regardless of mode, the high frequency cutoff can be used to reduce noise at the higher frequencies.

c. For those cases where a limited current span is to be monitored, a pair of high megohm resistors, available as the Model 6033, can be mounted internally within the Model 604. Although this method limits the current span to only four decades, it does slightly improve both the noise and the rise time characteristics over that obtainable with the multi-range Model 6041 Current Shunt. The compromise between noise and rise time still must be made by choosing the guarded or unguarded mode.

d. Low offset current minimizes zero offset with high source resistance, permitting maximum resolution when measuring current.

Control	Functional Description	Paragraph
VOLTS Switch	Selects full scale range Model 604 is to measure, or gain instrument is to amplify.	2-6
HIGH CUT HZ Switch	Varies the high frequency rolloff of the Model 604 and selects the maximum bandwidth to be measured. When the Switch is in the OFF Position the full bandwidth of the Model 604 is available.	2-5
METER Switch	Turns instrument off and on; disconnects meter; selects meter polarity; sets instrument for center zero operation.	2-2
Meter Pilot Light	Indicates instrument is on.	28
SUPPRESS Controls: FINE ADJ., MEDIUM, COARSE	Zeros amplifier output and determines the amount of zero suppression; up to ±1 volt dc between inputs.	2-6
ZERO CHECK Buttons	Shunts respective INPUT Receptacle. Locking either one allows single ended use. Locking both zeroes the meter. Opening both allows differential measurements.	2-2
GUARD-OFF Switch	Allows guarding for high speed measurements from high resistance sources when in GUARD Position.	2-2
(+) and (-) INPUT Receptacles.	Connect inputs to sources. May be used either singularly or conjointly for single ended or differential measure- ments respectively. Receptacles are Teflon-insulated triaxial connectors.	2-1

TABLE 2. Model 604 Front Panel Controls (Figure 1).

TABLE 3. Model 604 Rear Panel Controls (Figure 2).

Control	Functional Description	Paragraph
UNITY GAIN OUTPUTS: FROM (-) INPUT; FROM (+) INPUT	For use as an extremely linear preamplifier. Outputs are equal to respective inputs within $\pm 0.005$ % at dc, exclusive of offset, noise and drift.	2-11
GROUND Post	Connected to ground of all the outputs and the ground wire of the power cord.	
AMPLIFIER OUTPUT	For monitoring output.	2-11
LINE VOLTAGE Switch	Sets instrument for 117 volt or 234 volt operation.	2-2
FUSE	3AG Slow Blow. 117 volt operation: -1/8 ampere. 234 volt operation: -1/16 ampere.	2-8

TABLE 4. Model 6041 Differential Current Shunt Front Panel Controls (Figure 3).

Control	Functional Description	Paragraph
+INPUT Switch	Switches shunt resistor at input of the amplifier for the Model 604 +INPUT Receptacle, thus determining full scale current range Model 604 is to measure.	2-2
-INPUT Switch	Switches shunt resistor at input of the amplifier for the Model 604 -INPUT Receptacle, thus determining full scale current range Model 604 is to measure. Both switches are used in differential current measure- ments.	2-2
INPUT: + and -	Connects Model 6041 inputs to sources. May be used either singularly or conjointly for single-ended or differential measurements respectively. Receptacles are Teflon- insulated triaxial connectors.	2-2
TO 604 INPUTS: + and -	Connects Model 6041 to respective Model 604 Input Recep- tacles.	2-2
GND	Connected to ground of Models 604 and 6041.	

### SECTION 2. OPERATION

2-1. INPUT CONNECTIONS. The Model 604 has two input connectors, the +INPUT Receptacle and the -IN-PUT Receptacle, for use either single-endedly or for measuring the difference between two input signals.

a. The Model 604 INPUT Receptacles are Tefloninsulated triaxial connectors. The center terminal is the high impedance terminal; the inner shield is either guard or ground; the outer shield is ground. The inner shield of the INPUT Receptacle may be used as a driven guard with the GUARD-OFF Switch in the GUARD Position.

 $_{\rm b.}$  There are Keithley Accessories available which are designed to increase accuracy and convenience of input connections.

c. The Model 6011 Input Cable facilitates input connections. Table 5 indicates the color coding of the alligator clips. The center terminal is shielded by the inner braid of the triaxial cable up to the miniature alligator clip. If the unshielded clip causes pickup from nearby electric fields, remove it and connect the shielded lead directly to the source.

TABLE 5.Color Coding of AlligatorClips for Model 6011 Input Cable.

Lead	Circuit	Terminal
heavy wire with red clip cover	high	Center
thin wire with	5	Inner Shield
black clip cover thin wire with	guard	Outer Shield
green clip cover		

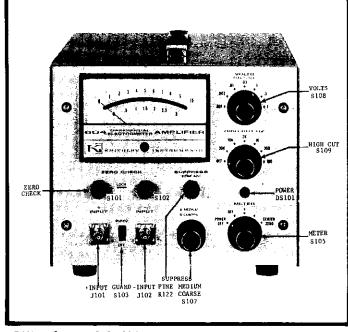


FIGURE 1. Model 604 Front Panel Controls

d. When working with a high impedance source carefully shield the input connection and the source since any variation in the electrostatic field near the input may cause definite meter disturbances.

e. Use high resistance, low-loss materials -such as Teflon (recommended), polyethylene or polystyrene -- for insulation. The insulation leakage resistance of test fixtures and leads should be several orders of magnitude higher than the internal resistance of the source. Excessive leakage reduces the accuracy of reading from high impedance sources. Triaxial or coaxial cables used should be a low noise type which employs a graphite or other conductive coating between the dielectric and the surrounding shield braid. Amphenol-Borg Electronics Corporation, Microdot, Inc., and Simplex Wire and Cable Company make satisfactory types. Use of the Model 6011 Input Cable insures good input connections.

#### NOTE

Clean, dry connections and cables are very important to maintain the value of all insulation materials. Even the best insulation will be compromised by dust, dirt, solder flux, films of oil or water vapor. A good cleaning agent is methyl alcohol, which dissolves most common dirt without chemically attacking the insulation. Air dry the cables or connections after washing with alcohol or use dry nitrogen if available. Or, if it is available, Freon is an excellent cleaning agent.

f. With the GUARD-OFF Switch in GUARD Position the Model 604 may have at least ten feet of triaxial cable on its input without adding capacitance to the

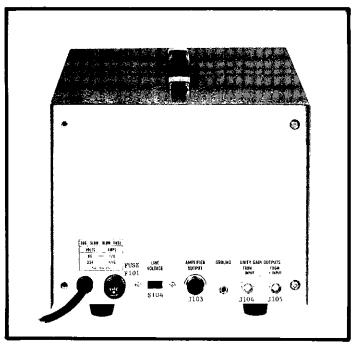


FIGURE 2. Model 604 Rear Panel Controls

input (input capacitance is specified at 1 pF). Note, however, that guarding can only eliminate input cable capacitance effects. Except in one special case, see subparagraph 2-li, guarding cannot be used to eliminate effects due to source capacitance. Do not connect the guard circuit to the source.

#### NOTE

For a complete discussion on guarding with the Model 604 refer to paragraphs 2-3 and 2-4.

9. When working with a high impedance source any change in the shunt capacitance of the input circuit will cause disturbances in the reading. Make the measuring setup as rigid as possible, and tie down connecting cables to prevent their movement. A continuous vibration may appear at the output as a sinusoidal signal, and other precautions may be necessary to isolate the instrument and the connecting cable from the vibration.

h. For low impedance measurements -- measurements that are above  $10^{-8}\ \text{ampere}\ \text{--}\ \text{unshielded}\ \text{leads}\ \text{may}$  be used.

i. The Model 6012 Triaxial-to-Coaxial Adapter enables using coaxial cables and accessories with the Model 604 by adapting the triaxial INPUT Receptacles to the UHF coaxial type.

1. The Adapter connects the Model 604 inner shield to ground defeating the guard capability that the triaxial receptacles make possible. Except for the special case spelled out in the following subparagraph 2, the GUARD-OFF Switch must be in the OFF position for the instrument to function.

2. If the Model 6012 Adapter is used with the Model 6041 Shunt in front of the Model 604 and the current source can be floated off ground, then a feedback picoammeter connection is possible. In this situation the coaxial shield is guard with the GUARD-OFF Switch in the GUARD position. To make possible a guarded circuit, connect guard, the coaxial shield, to the low of the current source. Remember, however, if the current source low is grounded, the GUARD-OFF Switch must be in the OFF position for the instrument to operate, and no guarded circuit is possible this way.

2-2. INPUT CONSIDERATIONS.

a. The Model 604 + INPUT Receptacle is the input to the non-inverting amplifier. This means that for a single-ended positive input to this Receptacle the output is positive and for a negative input the output is negative. The polarity can also be interpreted with the METER Switch and displayed on the meter.

b. The Model 604 - INPUT Receptacle is the input to the inverting amplifier. Thus for a singleended positive input signal to this Receptacle the output is negative and for a negative input the output is positive.

c. For single-ended measurements just lock the ZERO CHECK Button for the Input Receptacle that you do not intend to use, and apply the signal to the other Receptacle. When locked, the ZERO CHECK Button will connect its Input Receptacle to ground presenting an open circuit to the respective amplifier. The Model 604 always measures the signal differentially. In the single-ended mode it measures the difference between the applied signal and the reference signal (ground).

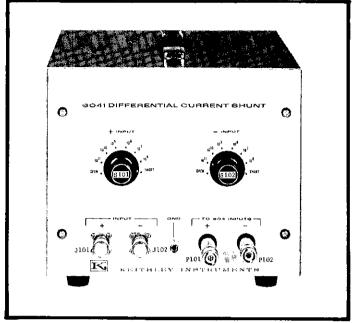


FIGURE 3. Model 6041 Front Panel Controls and Terminals.

d. For differential measurements unlock both ZERO CHECK Buttons and apply one signal to one of the Input Receptacles and the other signal to the other Receptacle. It does not matter which Receptacle accepts the high or low signal. If the signal to the -INPUT Receptacle is positive with respect to the signal to the +INPUT Receptacle, then the output, and the meter, will read negative. If the signal to the +INPUT Receptacle is positive with respect to the signal to the -INPUT Receptacle, then the output will be positive.

e. The Model 6041 is a Differential Current Shunt which, when used in conjunction with the Model 604, enables the Model 604 to become either a single-ended or differential shunt ammeter. The INPUT Receptacles on the Model 6041 are nominally labeled. That is, it is not essential that, single-endedly, a positive signal be applied to the (+) Receptacle and a negative to the (-) Receptacle or, differentially, the high signal to the (+) and the low to the (-).

1. The Model 6041 outputs (labeled: TO 604 INPUTS; + and -), which connect the Model 6041 to the Model 604, are directly tied to their corresponding Input Receptacle. That is, a signal to the +INPUT Receptacle will be accepted at the + output and a signal to the -INPUT Receptacle will be accepted at the - output.

2. Though it is not absolutely necessary, it is recommended that upon connecting the Model 6041 to the Model 604, the corresponding output receptacle of the Model 6041 be connected to the corresponding Input Receptacle on the Model 604. Otherwise, the user may become utterly confused at what the polarity at the Model 604 output corresponds to.

3. Note that the +INPUT Switch on the Model 6041 applies to the Model 6041 +INPUT Receptacle only and the -INPUT Switch applies to the -INPUT Receptacle only, regardless of the relative or absolute polarities of the signals at the inputs.

#### 2-3. GUARDING.

a. There are several factors which contribute to the bandwidth of an experiment.

- 1. The amplifier bandwidth;
- 2. The source resistance and capacitance;
- 3. The capacitance of the cable connecting the amplifier to the source.

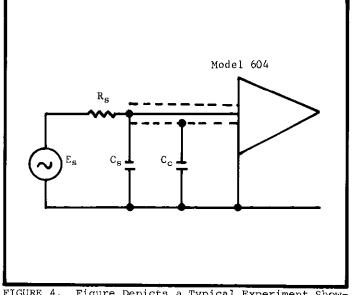


FIGURE 4. Figure Depicts a Typical Experiment Showing the Amplifier, Cable and Source. In the figure  $C_c$  is the capacitance due to the connecting cable,  $C_s$  is the source capacitance,  $R_s$  is the source resistance and  $E_s$  is the source voltage.

b. In the typical setup shown in Figure 4, if a perfect amplifier were placed at the source, the 3 dB frequency of the source would be

$$f_{3dB} = \frac{1}{2r} R_s C_s = .16 R_s C_s$$
 equation 1.

where  $f_{3dB}$  is the 3dB bandwidth of the source in Hz; R<sub>s</sub> is the source resistance in ohms. and C<sub>s</sub> is the source capacitance in farads.

1. If, to take an example,  $R_{\rm S}$  is equal to  $10^9$  ohms and  $C_{\rm S}$  is equal to 10 picofarads, then

$$f_{3dB} = \frac{.16}{(109)(10-11)} = \frac{.16}{(10-2)} = 16$$
 Hz.

2. Now, if a cable is used to connect the source to the amplifier, a new bandwidth is encountered because of the addition of the cable capacitance and equation 1 becomes

$$f_{3dB} = \frac{.16}{R_{g}(C_{g}+C_{c})}$$
 equation 2.

where  $f_{3dB}$  is the 3dB bandwidth of the system in  $_{\rm HZ\,2}$ 

 $R_{\rm S}$  is the source resistance in ohms;  $C_{\rm S}$  is the source capacitance in farads; and  $C_{\rm C}$  is the cable capacitance in farads.

An average cable adds approximately 30 picofarads per foot to the circuit. Thus, a 3-foot cable would add 90 picofarads and

$$f_{3dB} = \frac{.16}{109(10+90)(10-12)} = 1.6 \text{ Hz}.$$

So, in this example, connecting the source to the amplifier with a 3 foot cable, without guarding, would cause a 10:1 reduction in bandwidth.

c. The Model 604 is designed to eliminate (guard) the cable capacitance when used in the guard mode. The Model 604 will reduce the apparent capacity at the end of an up to 10 foot long triaxial cable to approximately 1 picofarad by driving the inner shield of the cable.

#### NOTE

The capacity from the center conductor of a triaxial cable to the inner shield is about 30 picofarads per foot. From the inner shield to the outer shield it is about 70 picofarads per foot. The signal source, however, only sees the capacity from the center conductor to the inner shield. The guard circuit, though, drives the entire capacity, about 100 picofarads per foot. The Model 604 guard circuit effectively drives up to 1000 pico-farads; thus the 10 foot cable limitation.

1. In the above example, then, the guard feature of the Model 604 allows the user to achieve almost the entire bandwidth of the source circuit.

$$f_{3dB} = \frac{.16}{10^9(10+1)(10^{12})} = 14.5 \text{ Hz}.$$

2. Notice that, in the above case, guarding does not increase the bandwidth of the source, but practically eliminates the effect of cable capacity.

d. The guard feature does, however, have several limitations.

1. It is effective driving only up to 1000 picofarads. Thus, no more than 1000 picofarads should be guarded from either input for optimum response.

2. The guard voltage swing is limited by the common mode voltage swing ( $\pm 11$  volts dc to 1 kHz and approximately  $\pm 2$  volts at 5 kHz).

3. The guard feature is not effective beyond 5 kHz because of internal phase shifts. These shifts cause peaking to occur above 5 kHz, but the system still remains stable.

e. Note that each input of the Model 604 has its own separate guard circuit which is completely independent of the other. Note, also, that guard and the unity-gain outputs are electronically identical (see schematic diagram 22820E in Section 5).

#### 2-4. GUARD-OFF SWITCH.

a. This Switch when in the GUARD Position reduces the effect of input cable capacity and provides high speed measurements from high resistance sources at the end of a cable by driving the cable capacitance and the inner shield of the INPUT Receptacle(s) (refer to paragraph 2-3). A guarded circuit is possible in this way.

b. With the Switch in the GUARD Position input capacity is decreased (to approximately 1 pF even with up to 10 feet of cable on the input) and rise time is decreased. Note, however, that the noise is increased.

c. With the GUARD-OFF Switch in the OFF position the input capacity is increased to 5 pF with no cable the input and 100 pF with 3 feet of cable on the input. The speed of response and the noise are decreased.

#### NOTE

When using the Model <u>6041</u> with the GUARD-OFF Switch in the GUARD position the total system is guarded. Also, less than 3 pF will be present at the Model <u>6041</u> input with 3 feet of cable on the input.

d. Therefore, although guarding increases speed, it also increases noise as well and a compromise between speed and noise will have to be made.

2-5. HIGH CUT HZ SWITCH.

a. This Switch allows the user to select the most amenable bandwidth to achieve optimum rise time and noise. The amplifier rolls off at 6 dB/octave and the Switch determines the measured bandwidth of the amplifier and cuts off higher frequencies at the 3 dB point. For example, if the Switch is set to 1 kHz the bandwidth of the Model 604 is 1 kHz and all higher frequencies are cut off. Setting the Switch to OFF gives the rated frequency response.

b. The user can use the HIGH CUT HZ Switch to select the optimum bandwidth for his measurement by cutting off higher frequency noise components, ringing and overshoot. Note, however, the lower the bandwidth the longer the rise time. So the user will have to use the requirements of his experiment and the method of trial and error to select the proper setting of the Switch for optimum achievement.

2-6. SUPPRESS CONTROLS.

a. There are three SUPPRESS Controls: FINE ADJ., MEDIUM and COARSE. The COARSE and MEDIUM SUPPRESS Controls are eight position switches. The MEDIUM Control interpolates between the setting of the COARSE Control. The FINE ADJ. Control is a tenturn potentiometer that interpolates between the settings of the MEDIUM Control. These controls may either be used for suppression or for zeroing the Model 604.

b. The zero suppress circuit cancels any constant dc voltage in order to use a more sensitive range to observe a superimposed signal. There is up to  $\pm 1$  volt available for zero suppression. This means that on the most sensitive ranges up to 1000 times full scale may be suppressed. For example, the Model 604 can measure changes of 1 millivolt full scale in a 1 volt steady signal on its .001 volt range.

#### c. Operation.

1. Adjust the VOLTS Switch to the range that gives the closest to a full scale meter deflection.

2. Turn the SUPPRESS Controls completely in the direction opposite meter deflection (counterclockwise for positive deflections and clockwise for negative deflections).

3. Turn the COARSE Control setting until the meter needle passes through zero. Turn the MED-IUM Control until the needle passes back through zero and then adjust the FINE ADJ. Control for zero deflection. 4. Set the VOLTS Switch to a more sensitive range and readjust for zero, if necessary.

2-7. COMMON MODE REJECTION.

a. A good differential amplifier is characterized by its common mode rejection and also its common mode voltage range. The Model 604 rates relatively well in both these areas.

b. The specifications in Table 1 read greater than 90 dB from dc to 120 Hz decreasing to 80 dB at 1 kHz. Actually, the common mode rejection is greater than this. The specification is based on the nonlinear distortion caused by swinging the amplifier between ±11 volts common mode. The fundamental is almost completely removed and all that appears at the output is predominately second harmonic of the common mode input signal. This distortion increases with frequency and the specification is degraded accordingly. At lower common mode levels, lets say 2 volts peak-to-peak, it is practically impossible to see any output change due to the common mode signal up to 1 kHz at any gain setting on the Model 604.

#### 2-8. PRELIMINARY PROCEDURES.

a. Check the Fuse and the LINE VOLTAGE Switch for the proper line voltage. Connect the power cord to the power source.

b. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
ZERO CHECK Button (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	POWER OFF.

c. Turn the METER Switch to CENTER ZERO. The Meter Pilot Light should turn on. Within a few seconds the meter needle should come to the center position. If not, adjust the meter zero with the FINE ADJ. and MEDIUM SUPPRESS Controls. Normally, there is no need to use the COARSE Control.

#### NOTE

Using the center zero scales decreases accuracy because the scale span is reduced.

d. After a few moments increase the voltage sensitivity by advancing the VOLTS Switch to .3, .1, etc. Continue zeroing with the FINE ADJ Control.

e. After long periods of storage or after an overload, the Model 604 may drift excessively. The input transistors are insensitive to mechanical shock; however, a severe input overload may cause a zero offset. This is corrected with the SUPPRESS Controls. Drifting, though, can occur for several hours.

#### 2-9. VOLTAGE MEASUREMENTS.

a. The Model 604 can be used to measure voltages two ways. Both ways may be used in either the guarded or the unguarded mode (refer to paragraphs 2-3 and 2-4).

1. In the single-ended method the unknown voltage is connected to one of the Input Receptacles. The other Receptacle is locked in Zero Check Position (refer to paragraph 2-2).

#### NOTE

The ZERO CHECK Buttons are true transfer type switches. When they are depressed, the input will be briefly connected to  $10^9$  ohms to ground. Then as the Buttons are further depressed toward LOCK position the input is open circuited and will remain so until the Button is released. In the LOCK position the Model 604 input is internally connected to ground. Please note that for certain very high impedance sources it may be necessary to never connect the input to ground, even through  $10^9$  ohms. If this is the case, depress and release the ZERO CHECK Button as fast as possible and the source will see  $10^9$  ohms for only a few milliseconds.

2. In the differential method one unknown voltage is connected to one of the Input Receptacles and the other voltage is connected to the other Input Receptacle. In this mode both ZERO CHECK Buttons are unlocked. (Refer to paragraph 2-2).

3. Accessory probes extend the Model 604's range to 10 kilovolts. (Use either single-endedly or differentially only with the GUARD-OFF Switch in the OFF position).

b. Single-Ended Method Voltage Measurements. This method should be used when an unknown voltage from a single source is to be measured. (Refer to paragraph 2-2 also).

1. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
ZERO CHECK Button (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	CENTER ZERO

2. Connect the unknown voltage to one of the Input Receptacles. Zero the meter (paragraph 2-6) and unlock the ZERO CHECK Button pertaining to the used Input Receptacle. Keep the other ZERO CHECK Button in the LOCK position. Set the METER Switch to + or -, as necessary. Increase sensitivity with the VOLTS Switch until the greatest on-scale meter deflection is obtained. Recheck zero setting after increasing sensitivity. (Refer to paragraph 2-2 also).

3. For guarded, fast measurements, set the GUARD-OFF Switch to GUARD. This method reduces the effects of input cable capacity with very high impedance sources and allows guarded voltage measurements (Refer to paragraphs 2-3 and 2-4).

4. Set the HIGH CUT HZ Switch to the desired position to obtain optimum response. (Refer to paragraph 2-5).

c. Differential Method Voltage Measurements. This method should be used to measure the difference between two unknown voltages neither of which has to be at ground potential, and either of which may be as much as  $\pm 11$  volts off ground. (Refer to paragraph 2-2).

1. Set the Model 604 front panel controls as shown in subparagraph bl above.

2. Connect one unknown voltage to one of the Input Receptacles and the other voltage to the other Input Receptacle. Zero the meter and then unlock both ZERO CHECK Buttons. Set the METER Switch to + or -, as necessary. Increase sensitivity with the VOLTS Switch until the greatest on scale meter deflection is obtained. Recheck zero setting after increasing sensitivity. The difference signal is equal to the percentage of full scale that the meter reads times the VOLTS Switch setting. (Refer to paragraph 2-2 also).

3. For guarded, fast measurements, set the GUARD-OFF Switch to GUARD. (Refer to paragraphs 2-3 and 2-4).

4. Set the HIGH CUT HZ Switch to the desired position to obtain optimum response. (Refer to paragraph  $2\sim5$ ).

d. To measure sources greater than 1 volt, use one of two divider probes. The Model 6102A 10:1 Divider Probe extends the Model 604's range to 10 volts; overall accuracy is ±4% and input resistance is 1010 ohms. The Model 6103A 1000:1 Divider Probe extends the Model 604's range to 1 kilovolt; overall accuracy is  $\pm 6$ % and input resistance is  $10^{12}$ ohms. Follow the same operating procedures with the dividers as in subparagraph b. The Model 6012 Triaxial-to-Coaxial Adapter must be used with the Models 6102A and 6103A Divider Probes. Note, however, using the Adapter connects the inner shield to ground, defeating the guarding capability of the Model 604. Therefore, the GUARD-OFF Switch must be in the OFF position for the instrument to operate. The full-scale voltage range is the divider ratio times the VOLTS Switch setting.

#### NOTE

If the Models 6102A and 6103A Divider Probes are used with the Model 604 in the differential mode, the common mode rejection is limited by the probe matching and typically would be about 30 dB. Thus, 30 volts of common mode voltage would cause a full scale indication.

2-10. CURRENT MEASUREMENTS. The Model 604 becomes an ammeter when used in conjunction with the Model 6041 Differential Current Shunt or when resistors are installed within the cases of the Model 301 (see Figure 6). When using the Model 6041, resistors are switched across the inputs of the Model 301 Amplifiers in the Model 604 with the + and - INPUT Switches on the Model 6041. The full scale current range is equal to reciprocal of the INPUT Switch setting on the Model 6041 times the setting of the VOLTS Switch on the Model 604. For example, if the INPUT Switch is set to  $10^8$  and the VOLTS Switch is set to .01, then the full scale current range is equal to  $1/108 \times .01 \approx 10^{-8} \times .01 \approx 10^{-10}$  amperes.

#### NOTE

The Model 301 Instruction Manual, supplied with the Model 604, gives complete information on Model 301 operation, circuit description, troubleshooting, calibration, parts lists and schematic diagrams. There are two Model 301 Operational Amplifiers used as plug-in units in the Model 604.

a. The Model 604 can measure currents several ways, each of which may be used either in the guarded or unguarded mode.

1. In the single-ended method the unknown current is connected to one of the Input Receptacles on the Model 6041. The other Input Receptacle on the Model 6041 is locked in Zero Check position



by its corresponding ZERO CHECK Button on the Model 604 front panel. (Refer to paragraph 2-2).

2. In the differential method one unknown current is connected to one of the Input Receptacles on the Model 6041 and the other current is connected to the other Input Receptacle. In this mode both ZERO CHECK Buttons are unlocked. (Refer to paragraph 2-2).

3. With the shunt resistors mounted inside the cases of the Model 301 Amplifiers in the Model 604 the Model 604 may be used as an ammeter either single-endedly or differentially. This method, however, limits versatility and gives only four decades of response. The same input considerations apply with this method as spelled out in paragraph 2-2.

#### CAUTION

When resistors are mounted internally, the input is open circuited when in zero check. This presents an open circuit to the current source, stopping current flow. In some cases this could be harmful to the current source and possibly to the Model 604 itself when high complicance voltages are encountered.

#### NOTE

When making measurements from high impedance sources or low current sources using the guarding feature, it may be desireable to set the HIGH CUT HZ Switch to the lowest setting, 100 Hz, to limit the noise bandwidth in some cases. The noise increase, when guarding, increases linearity with the bandwidth increase and the noise may eventually overload the amplifier. This would cause an error in a meter reading, but the error would go undetected unless the output was monitored on an oscilloscope.

b. Rise time varies primarily with the current

range, the input capacity and the method used (guarded or unguarded). See specifications, Table 1, page ii. On all ranges, the rise time in the guarded method is less than one second with the specified capacity across the input. Even with much larger capacities on the input the negative feedback maintains a relatively short rise time. Given a choice, it is better to place the Model 6041 nearer to the current source than to the data reading instrument. Transmitting the input signal through long cables decreases the responses speed and increases noise.

c. Single-Ended Method Current Measurements with Model 6041. This method is used to measure an unknown current from a single source (Refer to paragraph 2-2).

1. Connect the Model 6041 outputs, labeled TO 604 INPUT + and -, to the corresponding Input Receptacles on the Model 604 with the supplied mating cable. Set the Model 604 and 6041 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
ZERO CHECK Button (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	CENTER
+INPUT Switch	SHORT
-INPUT Switch	SHORT.

2. Connect the unknown current to one of the Input Receptacles on the Model 6041. Zero the meter and unlock the ZERO CHECK Button that corresponds to the Input Receptacle being used. Set the METER Switch to + or -, as necessary. Increase sensitivity with the VOLTS Switch and the Input Switch that corresponds to the Input Receptacle being used until the greatest on-scale meter deflection is obtained. Recheck zero setting after increasing sensitivity. (Refer to paragraph 2-2 also).

ZERO

3. The full-scale current range is the VOLTS

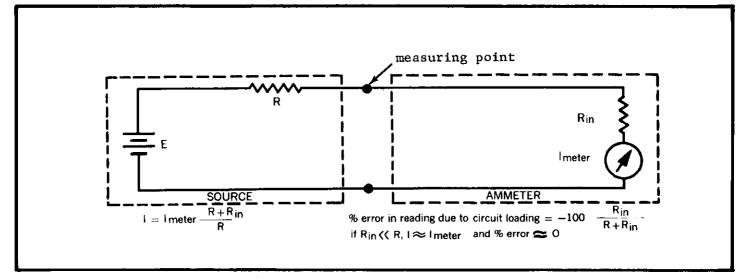


FIGURE 5. Error due to Ammeter Resistance. Current sources may be considered a voltage (E) in series with a resistance (R). The current with the ammeter short circuited is I = E/R. With the short circuit removed, the effective input resistance of the ammeter (R<sub>in</sub>) is in series with the source resistance (R). The current of the complete circuit is reduced and Imeter =  $E/(R+R_{in})$ . If the effective ammeter input resistance is small compared to R,  $I_{meter} \sim I$  and the error introduced by circuit loading is negligible.

Switch setting times the reciprocal of the Input Switch setting. Use the smallest VOLTS Switch setting possible to minimize input voltage drop and thus obtain the best accuracy. The full scale input voltage drop is equal to the VOLTS Switch setting.

#### NOTE

If the source has very high compliance voltage and large source resistance, the signal-to-noise ratio can be improved, without appreciably degrading the accuracy, by using a larger shunt resistor and a less sensitive voltage range. (For example: If a current source has a 300 volt compliance and  $10^{12}$  ohms output 300 volt compliance and  $10^{12}$  ohms output resistance, then I =  $300/10^{12} = 3 \times 10^{-10}$ ampere. Using a  $10^9$  ohms shunt resistor and a 300 millivolt full scale voltage range would display this current full scale. The loading error would be only 0.1%, which is 40 times less than the accuracy of the system. The signal-tonoise ratio would be 100 times better than if a  $10^7$  ohms shunt resistor and a 3 millivolt full scale range were used).

d. Differential Method Current Measurements with the Model 6041. This method is used to measure the difference between two unknown currents neither of which has to be at ground. (Refer to paragraph 2-2 also).

1. Connect the model 6041 outputs, labeled TO 604 INPUTS + and -, to the respective Input Receptacles on the Model 604 with the supplied mat-

ing cable. Set the Models 604 and 6041 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
ZERO CHECK Button (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	CENTER ZERO
+INPUT Switch	SHORT
-INPUT Switch	SHORT.

2. Connect one unknown current to one of the Input Receptacles on the Model 6041 and the other current to the other Input Receptacle. Zero the meter and unlock both ZERO CHECK Buttons. Set the METER Switch to + or -, as necessary. Increase sensitivity with the VOLTS Switch and both of the Input Switches until the greatest on-scale meter deflection is obtained. (Increase the settings of both Input Switches in unison). Recheck zero setting after increasing sensitivity. (Refer to paragraph 2-2).

3. The difference signal is equal to the percentage of full scale that the meter reads times the VOLTS Switch setting times the reciprocal of the setting of the Input Switches. Use the smallest VOLTS Switch setting possible to minimize input voltage drop and thus obtain the best accuracy. The full scale input voltage drop is equal to the VOLTS Switch setting.

e. Each Model 604 Input Receptacle is connected to a Model 301 Operational Amplifier hooked up in single-ended configuration. The Amplifier corresponding to the +INPUT Receptacle is in a non-inverting mode and the Amplifier corresponding to the -INPUT Receptacle is in an inverting mode. If so

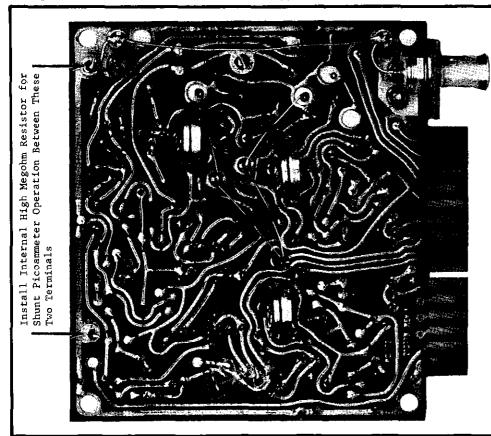


FIGURE 6. Internal View of Model 301 Amplifier Showing Location for Installing Internal Resistor. For complete and comprehensive information on the Model 301, refer to the supplied Model 301 Instruction Manual.

TABLE 6. Typical Rise Times and Noise of the Model 604 As An Ammeter With a Single Pair of Model 6033 Shunt Resistors Internally Mounted. This table was compiled using the smallest HIGH CUT HZ Switch setting permissable to obtain the rise times. Thus, the noise bandwidth of the system is kept as close as possible to the signal bandwidth.

Shunt	Current	100 pF Gua:	rded	100 pF Ung	uarded
Resistor	Span	Rise Time	Noise	Rise Time	Noise
(ohms)	(full_scale)	10%-90%	p-p	10%-90%	<u>p-p</u>
106 108 1010 1011	10-6 to $10^{-9}$ A 10-8 to $10^{-11}$ A 10-10 to $10^{-13}$ A 10-11 to $10^{-14}$ A	0.00003 sec 0.0003 sec 0.03 sec 0.3 sec	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	0.0003 sec 0.03 sec 3.0 sec 30.0 sec	$\begin{array}{c} 10^{-10} \text{ A} \\ 10^{-12} \text{ A} \\ 2\times 10^{-14} \text{ A} \\ 2\times 10^{-15} \text{ A} \end{array}$

desired, the Model 604 may be used as an ammeter by installing shunt resistors within the cases of either or both Model 301 Amplifiers. Figure 6 shows the positions in which the internal resistors may be mounted.

#### NOTE

When measuring current with the internally mounted Keithley Model 6033 Shunt Resistors, the Model 604 has slightly shorter rise times and lower noise for corresponding Model 6041 ranges. Please refer to Table 6.

1. In this configuration, therefore, the Model 6041 Current Shunt does not have to be used with the Model 604 to make current measurements. Note, however, that installing the resistors within the Model 301 cases is a time consuming operation. Note, also that using the Model 604 in such a manner severely limits the versatility and capability of the Model 604. Also, using it as an ammeter in such a way allows only four decades of current measuring capability.

#### NOTE

The Mcdel 604 may be used to measure voltages with the resistors installed within the Amplifiers. Note, however, that the magnitude of the input impedance would then be equal to the value of the installed resistor. This could lead to significant errors in voltage measurements if the source resistance is on the order of magnitude of the installed resistor.

2. The Model 604 may be used as either a single-ended or differential current measuring device in this configuration. The same instructions given in paragraph c and d above, less the Model 6041 instructions, apply here for the single-ended and differential modes respectively. The full scale current range, or difference signal, is equal to the reciprocal of the shunt resistor times the setting of the VOLTS Switch. (Refer to paragraph 2-2 also).

#### CAUTION

Pay special attention to the caution of paragraph 2-10a, sub-paragraph 3.

f. Use the GUARD-OFF Switch for Guarded or Unguarded Measurements. Guarded or unguarded measurements can be made in each of the preceeding modes. (Refer to paragraphs 2-3 and 2-4 also)

g. Use the HIGH CUT HZ Switch in each of the current moded as desired to obtain optimum response. (Refer to paragraph 2-5).

2-11. OUTPUTS. The Model 604 has three outputs:

two UNITY GAIN OUTPUTS each corresponding to its respective Input Receptacle, providing signals equal to each input-to-ground voltage to within ±0.005% at dc; an AMPLIFIER OUTPUT for recording the amplified or difference signal. These outputs may be used to drive recorders, oscilloscopes and X and Y recorders to record amplified signals, difference signals and absolute signals.

#### NOTE

The front panel SUPPRESS Controls will zero only the FROM +INPUT GAIN OUTPUT; that is the output for the non-inverting input. Also, the ±1 volt suppression may only be applied at this output. The FROM -INPUT UNITY GAIN OUTPUT may be zeroed internally by adjusting the zero control in the Amplifier for the -INPUT. (Refer to the supplied Model 301 Instruction Manual). Thus, when the AMPLIFIER OUTPUT is zeroed, it is highly unlikely that both UNITY GAIN OUTPUTS will be zeroed and several millivolts offset may be present. Also, it is very improbable that both the UNITY GAIN OUTPUTS and the AMPLIFIER OUTPUT will all be exactly zero at the same time, but any two of the three can be.

a. The AMPLIFIER OUTPUT is ±10 volts at 5 milliamperes to amplify signals within 1% for recorders. oscilloscopes and similar instruments. This output can be used for recording single-ended signals or differential, relative signals. The AMPLIFIER OUTPUT Receptacle is a microphone receptacle Amphenol type 80MC2F and mates with an Amphenol type 80MC2M. Pin No. 1 of the output terminal is the output and pin number 2 is ground when the GUARD-OFF Switch is in the OFF position.

b. The AMPLIFIER OUTPUT Receptacle and the UNITY GAIN Receptacles may be used in conjunction; the AMPLIFIER OUTPUT for recording the relative difference signal and a UNITY GAIN for recording the absolute signal of its input.

#### NOTE

All outputs of the Model 604 are short circuit proof.

c. The UNITY GAIN Outputs provide signals to each input-to-ground voltage within  $\pm 0.005$ % at dc. They can be used for impedance matching to minimize circuit loading errors or convenient connections to a recorder when the GUARD-OFF Switch is in the GUARD position. Also, these two outputs may be used to record with an XY recorder; 1 output to drive the X channel and one the Y.

### SECTION 3. CIRCUIT DESCRIPTION

#### 3-1. GENERAL.

a. The Model 604 is an entirely solid-state instrument that uses as its basic circuit components two Keithley Model 301 Operational Amplifiers. These two amplifiers, hooked-up in unity-gain configuration, are the input amplifiers for the Model 604. Three other major blocks of the Model 604 circuit are composed of three integrated circuit amplifiers that serve the functions of differential amplifier, auxiliary amplifier and an output amplifier, respectively. The 1st major block to be considered is the power supplies. Peripheral circuits result in zero suppression, frequency attenuation and amplifier gain through switched in resistors and capacitors.

b. The block diagram in Figure 7 illustrates the basic sections of the circuit of the Model 604. The schematic diagram, 22820E, is a detailed diagram showing the complete circuit of the Model 604 and the circuit designations. Refer to these to better understand the description given in the following paragraphs.

#### NOTE

The circuit description of the Model 301 Amplifiers, circuit designations AR101 and AR102, are not given in this particular section. However, the complete circuit description of these Amplifiers is in the supplied Model 301 Instruction Manual.

#### 3-2. INPUT AMPLIFIERS.

a. Two standard Keithley Model 301 Solid-State Electrometer Operational Amplifiers, designated AR101 and AR102, are used in the input; one for each input. They are connected for unity gain in such a fashion that an input resistance of greater than  $10^{14}$  ohms for the Model 604 is achieved. This is done by using the inverting input of the Model 301 as the input, connecting its output to ground and using the common as the output with the non-inverting input connected to common. In this configuration the gain accuracy is the reciprocal of the open loop gain of the Model 301 Amplifier.

b. Zero suppression is added between the non-inverting input and the common in the Model 301 Amplifier which is used in the Model 604 + input. Front panel controls Sl06 and Sl07 switch in resistors Rl22 through Rl44 to achieve this suppression.

#### NOTE

Refer to Table 7 for the gains of each amplifier in the Model 604 corresponding to each total gain of the Model 604.

3-3. DIFFERENTIAL AMPLIFIER. Circuit designation QA101 is a linear integrated circuit connected as a differential amplifier. The common mode rejection of the Model 604 is obtained by adjusting this amplifier, through capacitor Cl14 and resistor R148, both at low and high frequency.

3-4. AUXILIARY AMPLIFIER. Circuit designation QA102 is a linear integrated circuit connected for a non-inverting gain of 30. It is used only on the highest three gains, 1000 to 10,000 (refer to Table 7).

3-5. OUTPUT AMPLIFIER. Circuit designation QA103 is a linear integrated circuit connected as an inverting amplifier. Gain is adjustable between 1 and 33.3 (See Table 7). The high frequency rolloff

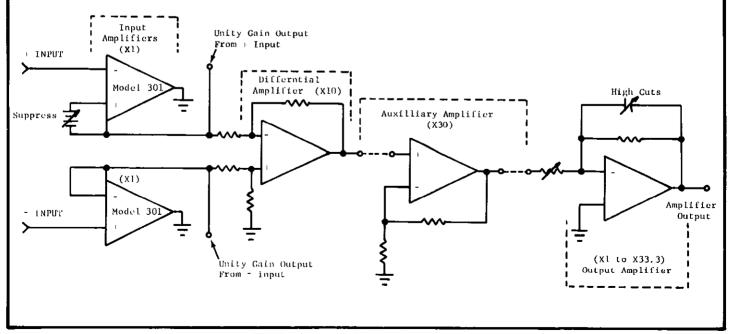


FIGURE 7. Model 604 Block Diagram.

#### TABLE 7. Model 604 Gain Chart

The Table gives the total gain of the Model 604 Electrometer Amplifier in one column, and the gain of each component amplifier within the Model 604 corresponding to each total gain. Refer also to Figure 7 which shows the basic amplifiers composing the Model 604 circuit.

Total Model 604 Gain	Input Amplifier Gain	Differential Amplifier Gain	Auxiliary Amplifier Gain	Output Amplifier Gain
10	1	10	<b></b>	
33.3	l	10	-	3.33
100	1	10	-	10
333	l	10	-	33.3
1000	1	10	30	3.33
3333	1	10	30	11.11
10,000	1	10	30	33.3

of this amplifier is adjusted between 100 Hz and wideband with the HIGH CUT HZ Switch, Sl08, by switching capacitors Cl21 through Cl26 across the feedback resistor, Rl53 through Rl57. The rolloff of these high cuts is 6 dB per octave. The output of QAl03 goes to output connector Jl03 and drives the meter, Ml01. By putting the HIGH CUT HZ Switch in the output amplifier, the noise bandwidth is always the same as the signal bandwidth.

3-6. POWER SUPPLIES. There are three sets of  $\pm 15$  volt power supplies in the Model 604. One set is located on the main board and drives the three linear integrated circuit amplifiers, QA101, QA102 and QA103. There is one set of supplies for each of the Model 301 Amplifiers located inside the Model 301s, with pre-regulators on the main board. (Refer to

supplied Model 301 Instruction Manual for a detailed description of Model 301 power supply regulator circuits). With the exception of the pre-regulators, the power supplies are all short circuit proof.

3-7. MODEL 6041 DIFFERENTIAL CURRENT SHUNT. The Model 6041 basically consists of two differentially connected, shielded, high megohm switches. Using either or both switches, high megohm resistors are shunted across the inputs of the respective Model 301 Amplifiers to make the Model 604 a differential or single-ended ammeter.

#### NOTE

Refer to schematic diagram 23458C for the Model 6041.

TABLE 8.

Equipment Recommended for Models 604 and 6041 Maintenance. Use these instruments or their equivalent.

Instruments	Refer to Paragraph
Ballantine Model 340 True RMS Voltmeter	4-8
Fairchild Hodel 7050 Digital Voltmeter	4-2,4-6,4-12,4-15,4-16
Hewlett-Packard Model 200CD Signal Generator	4-7,4-13,4-14
Hewlett-Packard Model 400 AC Voltmeter	4-13,4-14
Keithley Instruments Model 241 Voltage Supply	4-5
Keithley Instruments Model 260 Nanovolt Source	4-12,4-15,4-16
Keithley Instruments Model 261 Picoampere Source	4-16
Keithley Instruments Model 370 Recorder	4-11
Keithley Instruments Model 5155 High Megohm Resistor Standards	4-15
Keithley Instruments Model 6012 Triaxial-to-Coaxial Adapter	4-16
Keithley Instruments Model 610C Electrometer	4-2
Keithley Instruments Model 662 Differential Voltmeter	4-4,4-5
Keithley Instruments Type CS-178 Connector	4-2
Tektronix Model 561 Oscilloscope	4-3,4-4,4-7,4-13
Offset Current Fixture (composed of a 10 <sup>12</sup> ohm resistor and a Keithley Type CS-141 Connector)	4-9
100:1 Divider	4-7,4-13
1:1 and 1000:1 Dividers	4-13
10:1 Divider	4-13,4-14

### SECTION 4. MAINTENANCE

4-1. GENERAL.

a. The function of the maintenance section is to provide a method of checking the Models 604 and 6041 to make sure that they operate properly and within the specifications given in Table 1 on page ii.

b. The following procedures are recommended for adjusting and calibrating the Models 604 and 6041. Use the equipment recommended in Table 8. If proper facilities are not available or if difficulty is encountered, contact Keithley Instruments, Inc., or its representatives to arrange for factory calibration.

c. If returning the instrument to Keithley Instruments for repair or calibration, use the Repair and Calibration form at the rear of this manual. Use of this form will clarify difficulties that may ensue and will result in continued good service. When using this form, please detach along dotted line.

d. The figures at the end of this section show the component layouts for the Model 604. Refer to them to find desired components for adjusting the instrument. Refer also to the circuit description in Section 3 and the schematic diagram at the back of the manual to identify the more critical components.

#### NOTE

The supplied Model 301 Instruction Manual contains the troubleshooting, calibration procedures, component layouts and schematic diagrams for the AR101 and AR102 amplifiers.

						TABLE 9	).			
Mod	lel	60	4	Interr	al (	Controls.	. (Sec	Figures	at	end
of	thi	s	sc	ection	for	control	locatio	ons).		

Control	Circuit Designation	Paragraph
Zero Adjust (Model 301)	R106	4-4
+15V Adj.	R112	4-2,4-3
-15V Adj.	R120	4-2,4-3
Common Mode Adj.	R148	4-7
Common Mode Adj.	C114	4-7
Zero Center Adj.	R158	4-6
Full Scale Adj.	R161	4-6

#### 4-2. INITIAL CHECKS AND ADJUSTMENTS.

a. Make sure the Model 604 is not connected to an ac power line. Set rear panel LINE VOLTAGE Switch properly and the front panel controls as follows:

LOCK
LOCK
1
OFF
OFF
POWER OFF

b. Remove both Model 301 Amplifiers, circuit designation AR101 and AR102 (Figure 14), from the Model 604 chassis. Do not reinstall these ampli-

#### fiers until so specified in subparagraph 4-2i.

c. Adjust mechanical meter zero for zero indication on the meter.

d. Check for proper operation of both ZERO CHECK Buttons as delineated in subparagraphs 1 through 7 below.

#### NOTE

The ZERO CHECK Buttons are break-beforemake transfer switches with a  $10^9$  ohm resistor built into the shaft of the switch. This resistor places  $10^9$  ohms resistance to ground across the corresponding Model 301 Amplifier input (AR101 or AR102) during transfer between contacts. The  $10^9$  ohm resistance eliminates large transients during break.

 Set the ZERO CHECK Button for the + INPUT Receptacle to LOCK.

2. Connect the ARIO1 Model 301 Amplifier input (Figure 8) to a Model 610C Electrometer. Use an adapter or cable with a Keithley Type CS-178 jack on the end for connection to the ARIO1 Input. Operate the Model 610C in Normal mode on the 10 x  $10^9$  ohms range. With the + INPUT ZERO CHECK Button fully depressed or in LOCK, the Electrometer meter should indicate zero.

3. Release the ZERO CHECK Button slowly. As it is released the Model 610C should indicate approximately 1 x  $10^9$  ohms when the Button is about 1/16 to 1/64 of an inch from full depression. This point is the break from the ground contact.

4. The 1 x  $10^9$  ohms reading should continue until the ZERO CHECK Button nears the fully released position, at which time the reading should increase to greater than 10 x  $10^9$  ohms. This point is the make with the input contact when the  $10^9$  ohms resistance shaft moves away from the wiper.

5. Set the Model 610C for 10 x  $10^{10}$  ohms range to slow the reading. Depress the ZERO CHECK Eutton until a l x  $10^9$  ohms reading is achieved. This is the position at which the shaft of the Button touches the wiper. Note the distance that the shaft must be depressed to reach this point. The distance should be approximately 1/16 of an inch. This will ensure that the shaft will not readily touch the wiper and place  $10^9$  ohms across the input when the ZERO CHECK Button is in the unlocked position.

6. Release the ZERO CHECK Button and short the center pin of the + INPUT Receptacle to ground. The meter reading should go to zero.

7. Repeat steps 1 through 6 using the ZERO CHECK Button for the - INPUT Receptacle and the input jack for the AR102 Model 301 Amplifier.

e. Connect the Fairchild Model 7050 DVM between ground and the inter-shield of the Model 604 + IN-PUT Receptacle. Use the Model 7050 as an ohmmeter on the 150 kilohm range.

1. With the Model 604 GUARD-OFF Switch in the OFF position the Model 7050 should read zero.

With the Switch in the GUARD position the ohmmeter should read greater than 30 kilohms and typically from 50 to 70 kilohms.

2. Repeat the above tests with the Model 7050 connected between ground and inter-shield of the - INPUT Receptacle.

3. When completed set the Differential Amplifier METER Switch to OFF.

f. Plug the Model 604 into a 117 or 234 volt ac power source and turn the METER Switch to +, - or CENTER ZERO. (Make sure that the Model 301 Amplifiers, AR101 and AR102, are removed).

1. If the Differential Amplifier is operative, the meter reading will remain at or near zero.

2. If the instrument is inoperative, that is, if the meter pegs or indicates a large positive or negative up-scale reading, then set the +15 and -15 volt supplies per subparagraph g below. Do this before troubleshooting for other than an obvious problem.

g. Set the +15 volt regulated supply to approximately 15 volts with the +15V Adj. Potentiometer R112. In like manner set the -15 volt regulated supply to -15 volts with the -15V Adj. Potentiometer R120. (Figure 17 shows the locations of R112. and R120 and the layout of PC-195 in general).

h. Use a Model 610C to check for approximately 30 to 36 volts across capacitors Cl02(+), Cl03(-), Cl04(+) and Cl05(-); the unregulated positive and negative supply voltages Model 301 Amplifiers ARI01 and ARI02. Since the common for these supplies has no reference to ground, these voltages can be checked by connecting the Electrometer directly across each capacitor. Figure 16 shows the location of the capacitors.

NOTE

The voltage across these capacitors with the Model 301s installed will be approximately 29 volts. Hence, they are referred to as unregulated 29 volt supplies.

i. Set the Model 604 METER Switch to POWER OFF and install the Model 301 Amplifiers.

1. Set the Model 604 GUARD-OFF Switch to GUARD. Check for continuity between the inter-shield of the + INPUT Receptacle and the center pin of the UNITY GAIN OUTPUT, FROM + INPUT. In like manner check for continuity between the inter-shield of the - INPUT Receptacle and the center pin of the UNITY GAIN OUTPUT, FROM - INPUT.

2. If there is no continuity, check the respective Model 301 Amplifier for continuity between pin 2 and pin 15.

j. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
VOLTS Switch	l
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	ON

If it is not possible to bring the meter on scale or to zero it, then one of the following problems may be the cause. 1. The Model 301 Amplifiers may be defective. If this is so, then either they should be repaired or a new operational set must be substituted.

2. The Model 604 suppression may be defective. Check this out per paragraph 4-6.

3. The 21 volt regulated supplies may be defective. Refer to paragraph 4-3 to check the supplies.

4. If none of the above localizes the problem, then perform the Zero Balance and Unity Gain checks as shown in paragraphs 4-4 and 4-5.

4-3. POWER SUPPLY ADJUSTMENTS.

a. Connect the Tektronix Model 561 Oscilloscope between the Model 604 chassis and the proper test points shown in Figure 11 for performing the power supply checks.

b. Check the plus and minus 26 volt supplies. (During this check, the Model 604 should be connected to a 117 or 234 volt ac power source).

1. The voltage for this check should be 26 volts  $\pm 3$  volts.

2. The ripple must be less than 3 volts peak-to-peak.

3. Figure 8 shows a typical ripple obtained for the 26 volt supply.

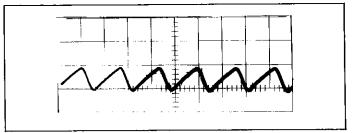


FIGURE 8. Typical Ripple Obtained for the 26 Volt Supply. The scale is 1 volt per division vertical, and 5 milliseconds per division horizontal.

c. Check the plus and minus 15 volt supplies with the Model 604 connected to a 117 or 234 volt ac power line.

 Adjust +15V Adj. Potentiometer, Rll2, for +15 volts ±30 millivolts. (Figure 17 shows Rll2).

 Adjust -15V Adj. Potentiometer, Rl20, for -15 volts ±30 millivolts. (Figure 17 shows Rl20).

3. The ripple in each case must be less than 2 millivolts peak-to-peak. Figure 9 shows typical ripple for the 15 volt supplies.

4. Monitor the plus and minus 15 volt supplies as the line voltage is changed from 105 to 125 volts ac for a 117 volt line or 210 to 250 volts for a 234 volt line. The voltage change in the 15 volt supplies should be less than 100 millivolts.

5. Return the line to 117 or 234 volts ac.

d. Check the regulated plus and minus 21 volt supplies of both Model 301s. Pin 6 and pin 10 of the Model 301s on the underside of the PC board are

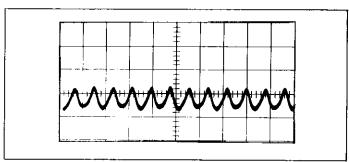


FIGURE 9. Typical Ripple Obtained for the 15 Volt Supply. The scale is 1 millivolt per division vertical, and 10 milliseconds per division horizontal.

the test points for the -21 volt supply and +21 volt supply respectively. Figure 15 shows the location of these test points.

1. The voltage at the positive and negative test points should be +21 and -21 volts  $\pm 2$  volts respectively.

2. The ripple in each case should be less than 40 millivolts peak-to-peak and is typically less than 30 millivolts peak-to-peak. Figure 10 shows a typical ripple for the 21 volt supplies.

NOTE

The Model 301 supply voltages are referenced to ground only when the 301s are installed. Use the common of capacitors Cl02 and Cl03 or Cl04 and Cl05 if the Model 301s are removed. Ripple measurements are meaningless with the Model 301s removed.

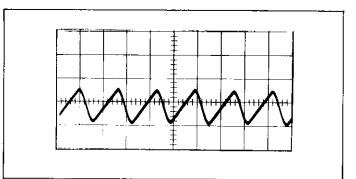


FIGURE 10. Typical Ripple Obtained for the 21 Volt Supply. The scale is 10 millivolts per division vertical, and 5 milliseconds per division horizontal.

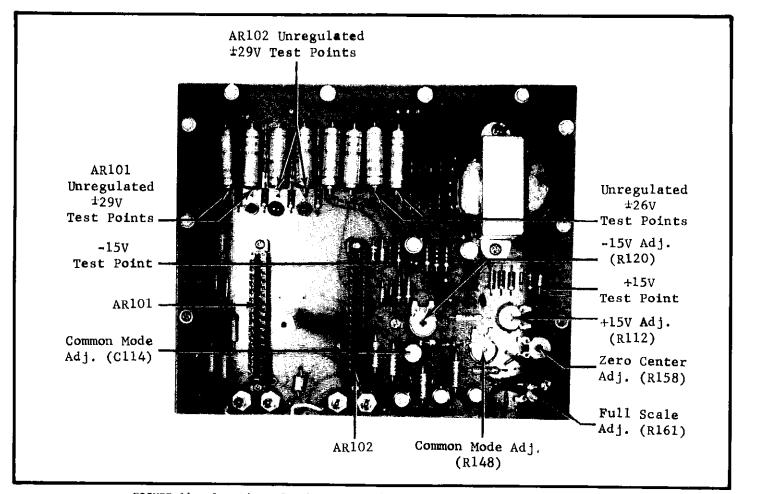


FIGURE 11. Location of Voltage Supplies Test Points (see paragraph 4-3).

4-4. MODEL 301 ZERO BALANCE ADJUSTMENT AND OSCILLATION CHECK.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
n n n (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	OFF

b. On the underside of PC-195, jumper the suppression voltage to low as shown in Figure 14. The suppression voltage is available at one input of the AR101 Model 301.

1. Connect a Keithley Model 662 Differential Voltmeter to the Model 604 UNITY GAIN OUTPUT, FROM + INPUT. Adjust the AR101 Model 301 Zero Control, available through the Model 301 cover, for zero volts ±1 millivolt.

2. Repeat the procedures in subparagraph 1 above using the UNITY GAIN OUTPUT, FROM - INPUT and the AR102 Model 301.

3. Remove the jumper attached in subparagraph 4-4b above. Set the Model 604 VOLTS Switch to .001 and zero the Meter with the Model 604 SUP-PRESS Controls. When completed return the VOLTS Switch Setting to 1.

c. Connect the UNITY GAIN OUTPUT, FROM + INPUT to an ac coupled Model 561 Oscilloscope and set it on the 1 millivolt per division vertical scale. Observe the Oscilloscope for high frequency oscillations as the Model 604 VOLTS Switch is stepped from 1 through .001 settings. If oscillations are observed, it can be corrected by replacing the 1.2 microfarad capacitor C103 in the Model 301.

NOTE

Do not mistake normal output noise for oscillation. Normal output noise is approximately 1 millivolt of grass and 120 cycle ripple.

d. Connect the UNITY GAIN OUTPUT, FROM - INPUT to the Oscilloscope and repeat the procedures of paragraph 4-4c above.

#### 4-5. UNITY GAIN CHECK.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
r r " (~)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	OFF

1. Connect a Keithley Model 241 Voltage Supply to the Model 604 + INPUT Receptacle and to the low terminal of the Keithley Model 662 Differential Voltmeter. Make sure, upon connection, that the Model 241 is dialed for zero volts and that the output is on. Ensure, also, that the link on the Model 662 is not attached between low and ground of the Differential Voltmeter,

2. Connect the Model 604 UNITY GAIN OUTPUT, FROM + INPUT Terminal to the high terminal of the Model 662. 3. Set the Model 662 dials to zero and the null switch to 10 millivolts. The Model 662 should indicate at or near zero.

#### NOTE

A large transient generated when the Model 241 output switch is charged from stand-by to on causes a zero shift which results in erroneous readings. The correct method to use is to dial out zero volts then turn the output switch to on.

b. With the Model 241 at zero volts, release the Model 604 ZERO CHECK Button for the + INPUT Receptacle and note that the Model 662 null remains the same.

1. Apply a positive 10 volts with the Model 241. The Model 662 null indication should not change more than 0.5 millivolts and typically is less than 0.2 millivolts. A change of 0.5 millivolts indicates a unity gain of 50 ppm.

2. Return the Model 241 to zero volts. The Model 662 null should return to the original reading.

c. Repeat the procedures in above paragraphs a and b with the Model 241 connected to the Model 604 - INPUT Receptacle and the Model 604 UNITY GAIN OUT-PUT, FROM - INPUT Terminal connected to the high terminal of the Model 662.

4-6. SUPPRESSION AND METER CALIBRATION.

a. Set the Model 604 front panel contacts as follows:

ZERO CHECK Button	(+)	LOCK
	(-)	LOCK
VOLTS Switch		1
HIGH CUT HZ Switch	1	OFF
GUARD-OFF Switch		OFF
METER Switch		OFF

b. Connect the Model 604 AMPLIFIER OUTPUT to the Model 7050 Digital Voltmeter.

c. Check to make sure that the Model 604 SUPPRESS Controls are able to adjust the output for greater than plus and minus 10 volts. Then set the output for exactly 10.000 volts (10.000 volts is achieved when the Model 7050 alternates between 9.99 and 10.00).

d. Adjust the Full Scale Potentiometer, R161 (Figure 17), for full scale meter indication.

e. Set the VOLTS Switch to .001 and zero the Model 604 output and meter.

1. Set the VOLTS Switch to 1 and adjust the FINE ADJ SUPPRESS Control for exactly 0.000 volts at the output, if necessary.

2. Set the METER Switch to CENTER ZERO and adjust the Zero Center Potentiometer, R158 (Figure 17), to obtain a zero meter reading on the center zero scale.

4-7. COMMON MODE REJECTION CALIBRATION.

a. Set the Model 604 front panel controls as follows:

ZERO	CHECK	BUTTON	(+)	LOCK
	п		(-)	LOCK

VOLTS Switch	.001
HIGH CUT HZ Switch	100
GUARD-OFF Switch	OFF
METER Switch	OFF

b. Zero the Model 604 meter with the SUPPRESS Controls.

c. Connect the Hewlett-Packard Model 200CD Signal Generator to both the + INPUT Receptacle and -INPUT Receptacle of the Model 604 and apply 5 Hz at 20 volts peak-to-peak. Connect the Model 604 AMPLI-FIER OUTPUT to a dc coupled Model 561 Oscilloscope.

1. Release both ZERO CHECK Buttons simultaneously and adjust the Common Mode Adj. Potentiometer, R148 (Figure 17) for minimum output. Minimum output should occur at a null (a null is achieved where the voltage increases when the potentiometer is adjusted either side of minimum output).

2. The output voltage must be less than 6 volts peak-to-peak (this is 90 dB of rejection at the amplifier gain of  $10^4$ ). Typically the minimum output will be less than 3 volts peak-to-peak and will be composed basically of second harmonic.

3. When completed with the above tests, zero check both Model 604 INPUT Receptacles.

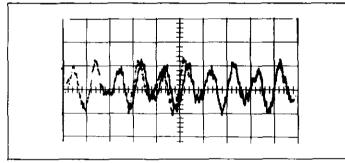


FIGURE 12. Null Obtained with the Signal Generator at 5 Hz. The scale is 1 volt per division vertical, and 0.1 second per division horizontal.

d. Set the Model 604 HIGH CUT HZ Switch to 10K and apply 1 kilohertz at 20 volts peak-to-peak to the Model 604 with the Model 200CD Signal Generator. Zero the meter with the SUPPRESS Controls.

1. Release both ZERO CHECK Buttons simultaneously and adjust the Common Mode Adj. Trimmer capacitor, Cll4 (Figure 16), for minimum output null. If null is unatainable, increase or decrease the value of capacitor Cl27 (Figure 16) as needed.

2. The output voltage must be less than 20 volts peak-to-peak (this is 80 dB of rejection at the amplifier gain of  $10^4$ ). Typically the mini-mum output will be less than 8 volts peak-to-peak and will be composed basically of second harmonic.

3. When completed with the above tests, zero check both Model 604 INPUT Receptacles.

e. If, in either or both of the preceding checks of paragraphs c and d, the minimum output exceeds the stated limits, then the probable cause is high second harmonic generated by integrated circuit amplifier QA101 (Figure 14). If this is so, re-

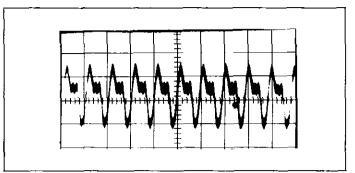


FIGURE 13. Null Obtained with the Signal Generator at 1 kHz. The scale is 2 volts per division vertical, and 1 millisecond per division horizontal.

place QA101 with a new component.

f. Set the HIGH CUT HZ Switch to OFF and set the Signal Generator to 75 kilohertz at minimum output. Connect a 100:1 divider between the Signal Generator and the + and - INPUT Receptacles of the Model 604. Connect an ac coupled Model 561 Oscilloscope to both Model 604 UNITY GAIN OUTPUT Receptacles. Set the Oscilloscope to 50 millivolts per division vertical. It may be necessary to float the Oscilloscope to eliminate 60 cycle ground loops.

#### NOTE

Monitoring the UNITY GAIN OUTPUT Terminals is effectively the same as monitoring the input.

1. Set the METER Switch to on (+, - or CENTER ZERO). Zero the Model 604 meter with the SUPPRESS Controls and release both ZERO CHECK Buttons simultaneously.

2. Observe the meter and slowly increase the Signal Generator output until a dc shift is observed.

3. Decrease the signal to the point where the dc shift just starts to occur.

4. The signal at the input, monitored at the UNITY GAIN OUTPUT Terminals with the Oscilloscope, should be 100 millivolts peak-to-peak or greater.

#### NOTE

For the calibration procedures in paragraph 4-8 through 4-15, it is imperative that the Model 604 cover be on the unit.

4-8. NOISE CHECK.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
"""(-)	LOCK
VOLTS Switch	.001
HIGH CUT HZ Switch	100
GUARD-OFF Switch	OFF
METER Switch	CENTER ZERO

b. Attach the Ballantine Model 340 True RMS Voltmeter to the Model 604 AMPLIFIER OUTPUT Receptacle.

1. Zero the Model 604 with the SUPPRESS Controls.

2. Observe the output noise and meter noise.

a) The meter noise, excluding drift, should be less than 80 millivolts peak-to-peak.

b) The output noise, excluding drift, must be less than 200 millivolts RMS.

3. Set the GUARD-OFF Switch to the GUARD position and check the noise. There should be no appreciable change in noise. Return the Switch to OFF.

c. Set the HIGH CUT HZ Switch to OFF and observe the output noise. It should be less than 600 millivolts RMS.

4-9. OFFSET CURRENT CHECK. This check requires a special, easy-to-construct, fixture. The fixture consists of a triaxial connector, Keithley part number CS-141, and a shielded  $10^{12}$  ohm resistor that connects the center pin of CS-141 to ground. The fixture must be shielded to eliminate excessive noise pickup.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button	(+)	LOCK	
n n n	(-)	LOCK	
VOLTS Switch		.001	
HIGH CUT HZ Switch	1 I	OFF	
GUARD-OFF Switch		OFF	
METER Switch		CENTER	ZERO

b. Connect the Current Offset Fixture to the Model 604 + INPUT Receptacle and zero the meter.

1. Set the VOLTS Switch to .1. In conjunction with the  $10^{12}$  ohm resistor in the Fixture, this results in  $10^{-13}$  ampere full scale.

2. Release the + ZERO CHECK Button and note the offset current. It should be less than 2 x  $10^{-14}$  ampere (this is within 2 major divisions of zero). The offset may be positive or negative.

c. Repeat the procedure of above paragraph b with the Fixture on the - INPUT Receptacle.

4-10. 234 VOLT AC OPERATION CHECK. (For 117V units).

a. Set the LINE VOLTAGE Switch on the rear panel of the Model 604 to 234V and plug the unit into a 234 volt ac 50 or 60 Hz power source.

b. Set the VOLTS Switch to .001 and Zero the meter with the SUPPRESS Controls. Being able to zero on this range is sufficient indication that the instrument is operative at 234 volts ac.

c. Unplug the instrument from the 234 volt ac power source and return the LINE VOLTAGE Switch to 117.

4-11. DRIFT VERIFICATION.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	LOCK
" " ()	LOCK
VOLTS Switch	.01
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	OFF

b. Connect the Model 604 AMPLIFIER OUTPUT Receptacle to the Keithley Model 370 Recorder. Set the Recorder attenuator to 10 volts. Recorder calibra-

# tion is now 10 millivolts full scale either side of zero.

c. After a one hour warm-up, the Model 604 must not drift more than 4 millivolts per week or more than 0.3 millivolts per  $^{\rm OC}$ .

d. While the drift specifications is in terms of a week, it may be possible to determine the drift within 24, 48 or 72 hours rather than run the full week. After the one hour warm-up and excluding drift due to temperature, the drift may be calculated as shown in the subparagraphs below.

1. After 24 hours, if the drift is less than 0.55 millivolt, the instrument meets specifica-tions.

2. After 48 hours, if the drift during the last 24 hours is multiplied by six and is added to the drift during the first 24 hours and the total is 4 millivolts or less, then the Model 604 meets specifications.

3. After 72 hours, if the drift during the last 24 hours multiplied by 5 is added to the drift of the first and second 24 hours and the total is 4 millivolts or less, then the unit meets the week's drift specifications.

e. Since temperature can be a major contributor to drift (up to 0.3 millivolt per  $^{\rm OC}$  allowable), the temperature should be recorded along with the drift.

f. If the drift of the Model 604 is excessive, age the instrument and redrift. If it is still excessive, then a probable cause of poor Model 604 drift is excessive Model 301 Amplifier drift. One or both of the Model 301s may have excessive drift. Or, it may be that the drift of that particular pair of Model 301s is additive, in which case replacing one Model 301 might solve the problem.

4-12. VOLTAGE ACCURACY VERIFICATION.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	Unlock
и и и (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	OFF

b. Connect the Keithley Model 260 Nanovolt Source to the Model 604 + INPUT Receptacle. Monitor the Model 604 AMPLIFIER OUTPUT Receptacle with the Model 7050 DVM.

c. Starting with setting of 1, check all the settings of the VOLTS Switch down to .001 for 10 volts ±1% at the output. This is an indication of from 9.90 volts to 10.10 volts on the Model 7050.

4-13. FREQUENCY RESPONSE CHECK.

a. Set the Model 604 front panel controls as follows:

(+)	LOCK
(-)	LOCK
	1
1	$\mathbf{OFF}$
	$\mathbf{OFF}$
	OFF
	(-)

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b. Connect a Signal Generator through a 1:1 divider to both the + INPUT and - INPUT Receptacles of the Model 604. Monitor the AMPLIFIER OUTPUT Receptacle with both a Hewlett-Packard Model 400 AC Voltmeter and an ac coupled Model 561 Oscilloscope. Use the Voltmeter on the 10 volt range and the Oscilloscope should be at 5 volts per division vertical.

NOTE

If other than a constant amplitude signal generator is used, it will be necessary to monitor the output of the signal generator with another voltmeter in order to maintain the same signal level at the various frequencies. Do not use the Model 604 UNITY GAIN\_OUTPUT Terminals for monitoring.

c. Set the Signal Generator for 100 Hz.

1. Unlock the + ZERO CHECK Button and adjust the Signal Generator amplitude to obtain 7 volts RMS at the output of the Model 604. The signal observed on the Oscilloscope should be approximately 20 volts peak-to-peak with no discernible distortion.

2. Sweep the frequency from 100 Hz to 10 kilohertz. The output must remain within 1% of the 100 Hz 7 volt RMS reference (6.9 volts to 7.1 volts RMS) and the signal observed on the Oscilloscope should have no discernible distortion.

3. Set the + ZERO CHECK Button to LOCK and unlock the - ZERO CHECK Button. The output signal at 10 kHz must be 7 volts RMS ±1% and the signal observed on the Oscilloscope should have no discernible distortion.

4. Set both ZERO CHECK Buttons to LOCK.

d. Replace the 1:1 divider with a 10:1 divider. Set the Signal Generator to 75 kHz and the Oscilloscope to 1 volt per division. Ignore the Voltmeter for this test.

1. Open the + INPUT Receptacle and note that the output will swing at least 2 volts peak-to-peak without discernible distortion.

2. Zero check the + INPUT Receptacle and repeat the procedure of subparagraph 1 above with the - INPUT Receptacle open.

3. Set both ZERO CHECK Buttons to LOCK.

e. Replace the 10:1 divider with a 100:1 divider. Set the Signal Generator to 10 kHz and the Voltmeter to the 100 millivolts range. Ignore the Oscilloscope for this test.

1. Unlock the + ZERO CHECK Button and adjust the Signal Generator amplitude to obtain 70 milli-volts at the Model 604 AMPLIFIER OUTPUT.

2. Sweep the frequency from 10 kHz to 200 kHz. The output must remain with  $\pm 3$  dB of the 10 kHz 70 millivolts reference.

3. Set the + ZERO CHECK Button to LOCK position and unlock the - ZERO CHECK Button. Again, the 200 kHz output signal must be within  $\pm 3$  dB of the 10 kHz 70 millivolts reference.

4. Set both ZERO CHECK Buttons to LOCK.

f. Set the Model 604 VOLTS Switch to .001 and

zero the meter. The meter must be maintained at or near zero during the checks in the following subparagraphs g and h or the output signal will be clipped.

g. Set the Voltmeter to the 10 volt range and the Oscilloscope to 5 volts per division vertical. Use a 1000:1 divider and check the .001 setting of the Model 604 VOLTS Switch per above paragraph 4-13c. (Note that there will be noise riding on the signal observed on the Oscilloscope).

h. Use a 1000:1 divider and set the Signal Generator to 10 kHz, the Voltmeter to the 3 volt range and the Oscilloscope to 2 volts per division.

1. Unlock the + ZERO CHECK Button and adjust the Signal Generator amplitude to obtain 2 volts RMS at the output of the Model 604. The signal observed on the Oscilloscope should be approximately 6 volts peak-to-peak and noisy.

2. Sweep the frequency from 10 kHz to 50 kHz. The output should remain within  $\pm 3$  dB of the 10 kHz 2 volts RMS reference.

3. Set the + ZERO CHECK Button to LOCK and unlock the  $\sim$  ZERO CHECK Button. Again, the 50 kHz output signal must be within ±3 dB of the 10 kHz 2 volts RMS reference.

4. Zero check both inputs of the Model 604.

4-14. HIGH CUTS CHECK.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	Unlock
е н н (~)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	OFF
METER Switch	OFF

b. Connect the Signal Generator through a 10:1 divider to the Model 604 + INPUT Receptacle. Connect the Model 604 AMPLIFIER OUTPUT Receptacle to the Model 400 AC Voltmeter and set the Voltmeter to the 3 volt range.

c. With the HIGH CUT HZ Switch at OFF, set the Signal Generator to 100 kHz and adjust the amplitude to obtain 2 volts RMS at the output of the Model 604. This level will be used as a reference throughout the High Cuts checks (paragraphs d and e below).

NOTE

Refer to note of paragraph 4-13b concerning the use of a constant amplitude signal generator.

TABLE 10.						
High	Cut	Ηz	Settings	for	Frequency	Check.

HIGH CUT Hz Switch Setting	3 dB Down Frequency Limits
100	80 to 120 Hz
300	240 to 360 Hz
1K	800 to 1200 Hz
3к	2.4 to 3.6 kHz
10K	8 to 12 kHz
30K	24 to 36 kHz

d. Set the HIGH CUT HZ Switch to the settings indicated in Table 10. Check that the output is 3 dB down from the 2 volts RMS reference at some frequency within the frequency limits listed for each setting.

e. If the output does not roll off at 3 dB within the frequency limits listed, the applicable high cut capacitor is out of tolerance.

4-15. INPUT IMPEDANCE CHECK.

a. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	Unlock LOCK
VOLTS Switch HIGH CUT HZ Switch GUARD-OFF Switch	1 OFF OFF
METER Switch	OFF

b. Connect the Model 260 Nanovolt Source to the Model 604 + INPUT Receptacle and the Model 7050 DVM to the AMPLIFIER OUTPUT Receptacle.

1. Apply + 1 volt to the + INPUT Receptacle. The output voltage should be 10 volts ±1% (9.90 to 10.10 volts).

2. Connect a Keithley Model 5155 High Megohm Standard  $10^{12}$  ohms resistor in series between the Model 260 and the + INPUT Receptacle.

3. Apply + 1 volt to the + INPUT Receptacle. The output voltage must come to within 100 millivolts of the voltage noted in subparagraph 1 above. This indicates an input impedance of  $10^{14}$ ohms or greater. Note, however, that it may require a few minutes for the reading to come in and stabilize.

c. Repeat the procedure in paragraph b using -l volt and the - INPUT Receptacle.

4-16. MODEL 6041 DIFFERENTIAL CURRENT SHUNT CALI-BRATION.

a. Check the full scale accuracy of the Model 604 on the 1 volt range per paragraph 4-12. Note the actual output voltage and use this as a reference to detormine the accuracy of the Model 6041 ranges. b. Set the equipment up in feedback picoammeter configuration as specified in the following subparagraphs.

1. Set the Model 6041 + INPUT and - INPUT Switches to the 10<sup>6</sup> position. Attach the TO 604 INPUTS + and - Connectors to their respective Model 604 INPUT Receptacles.

2. Set the Model 604 front panel controls as follows:

ZERO CHECK Button (+)	Unlock
n n n (-)	LOCK
VOLTS Switch	1
HIGH CUT HZ Switch	OFF
GUARD-OFF Switch	GUARD
METER Switch	OFF

3. Connect a Model 7050 DVM to the Model 604 AMPLIFIER OUTPUT Receptacle.

4. Remove the link from between low and ground on the rear of the Keithley Model 261 Picoampere Source. Connect the Model 261 to the Model 6041 + INPUT Receptacle through a Keithley Model 6012 Coaxial-to-Triaxial Adapter.

c. Check the full scale accuracy of the  $10^6$  through  $10^{11}$  settings of the + INPUT Switch with the Model 261 set for 1 x  $10^{-6}$  through 1 x  $10^{-11}$  ampere. All ranges must be within ±2% of the Model 604 full scale accuracy (actual output voltage ±0.20 volt).

1. After checking the accuracy set the Model 6041 + INPUT Switch to 10<sup>11</sup> and the Model 604 + ZERO CHECK Button to LOCK and the GUARD-OFF Switch to OFF. This places the equipment in shunt plcoammeter configuration.

2. Release the + ZERO CHECK Button and note that the meter reading climbs slowly to approximately 1/2 scale indication.

3. When completed, set the + ZERO CHECK Button to LOCK and the GUARD-OFF Switch to GUARD.

d. Repeat the procedure of paragraph c above with the Model 261 connected to the Model 6041 - INPUT Receptacle.

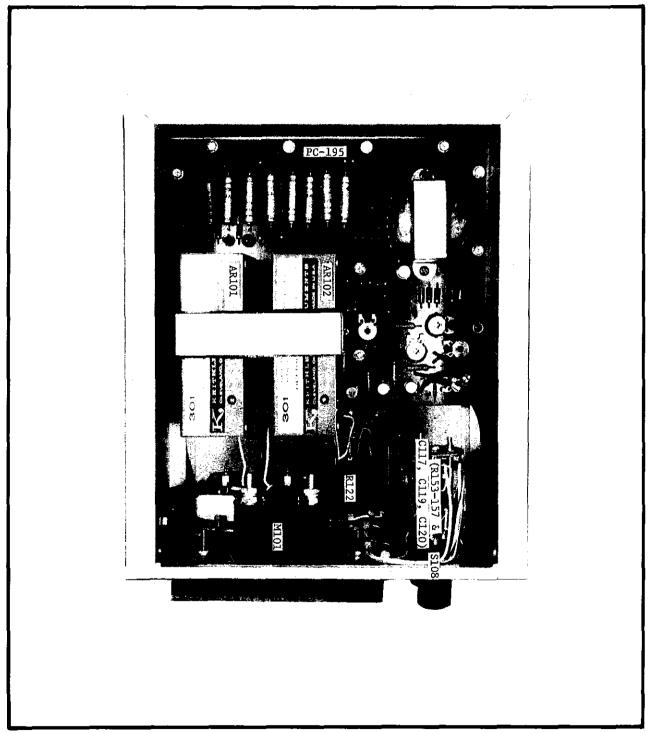
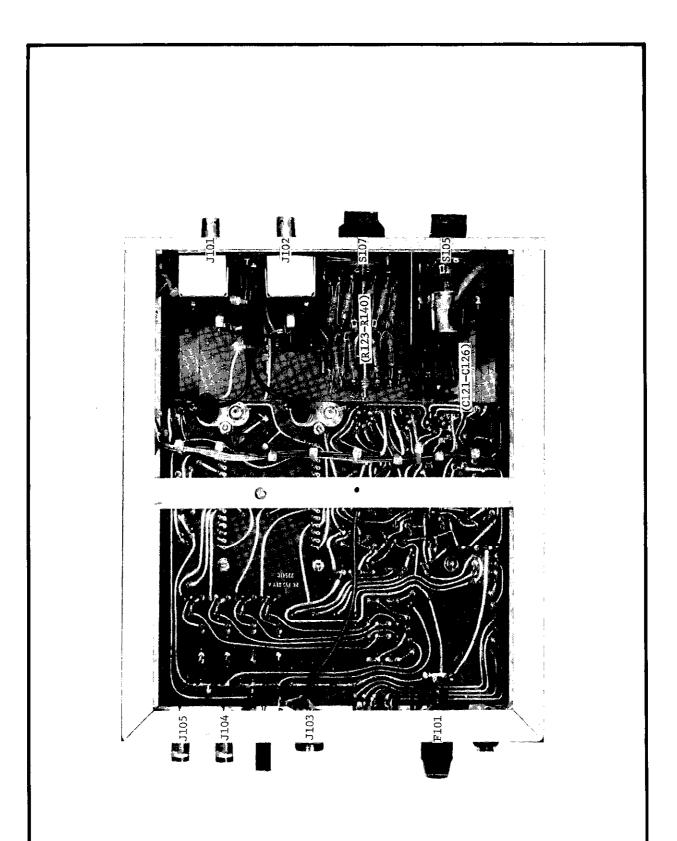


FIGURE 14. Top View Model 604 Chassis. Front panel faces up. Figure shows location of components, pc boards, and switches. For Bottom View, see Figure 15.



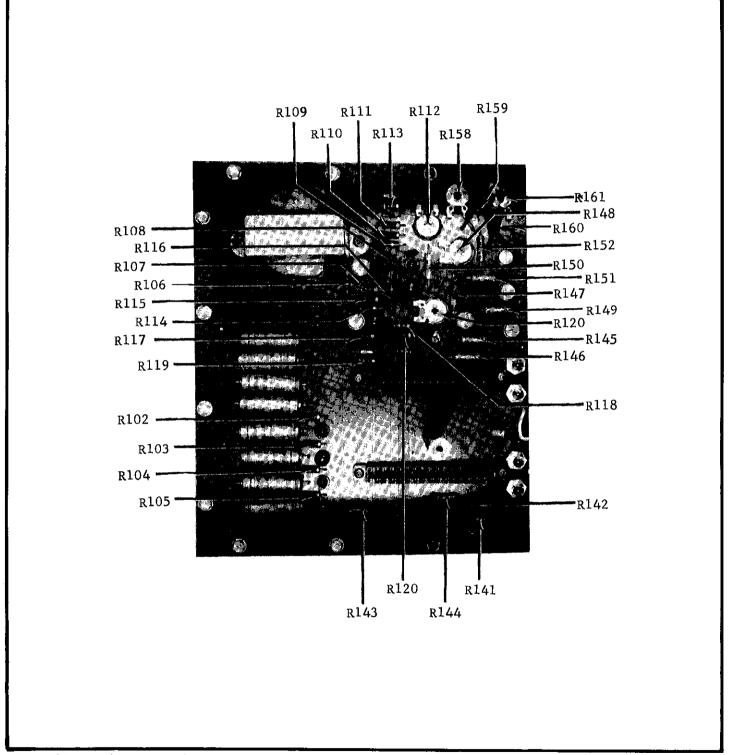
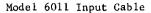


FIGURE 17. Resistor Locations on PC-195. For other component locations, see Figure 16.

## SECTION 5. ACCESSORIES

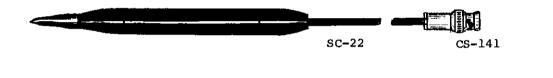


Description: The 6011 is a low-noise triaxial cable, 30" long, terminated by three color-coded alligator clips. This cable mates directly with the triaxial input. The cable is fabricated using a Keithley part no. CS-141 connector and part no. SC-22 low-noise cable. Application: The 6011 may be used for measurements which require a triaxial connection, especially when the input LO is floated above CASE ground. The cable permits full use of the Model 615; capabilities.





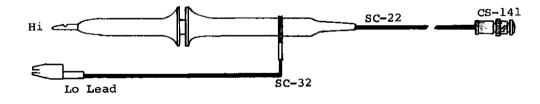
Description: The 6301 is a guarded triaxial cable, 3 tt. long, terminated by a probe for making point-topoint measurements. Application: The 6301 may be used for measurements which require a triaxial cable with a guarded probe having an insulation resistance greater than  $10^{14}$  ohms.



Model 1531 Gripping Probe

Description: The 1531 is a triaxial cable, 3 ft. long, terminated by a special gripping-type probe. The 1531 insulation resistance is greater than  $10^{10}$  ohms. The probe is rated for off ground measurements up to 500 v

Application: The 1531 may be used for measurements which require a triaxial cable. The probe permits convenient connections to the circuit under test due to the gripping feature.



#### Model 6012 Triax-to-Coax Adapter

Description: The 6012 is an adapter for mating the triaxial input and UHF (coax) type connectors. This adapter can be used with Models 6101A, 6101B, 6103A, 6102A, etc.



#### Model 4006 Rack Mounting Kit

### Description:

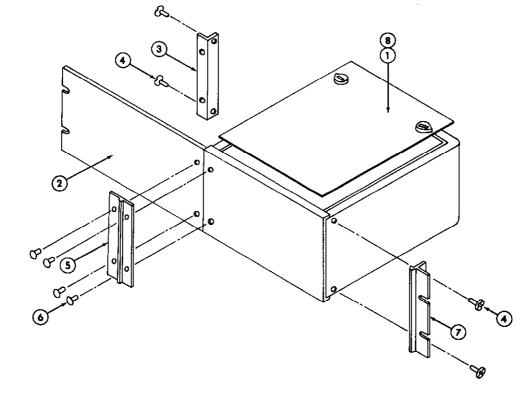
The Model 4006 is a rack mounting kit with overall dimensions, 7 inches high x 19 inches wide. Two top covers are provided for use with either 10 inch or 13 inch deep instruments.

### Application:

The Model 4006 converts the instrument from bench mounting to rack mounting. It is suitable for mounting one instrument in one-half of a standard 19-inch rack.

Parts	List:	
-------	-------	--

Item		Qty. Per	Keithley
No.	Description	Assembly	Part No.
1	Top Cover, 10"	1	200168
2	Panel Adapter Plate	1	19158A
3	Angle Support	1	19157A
4	Screw, #10 x 3/8"	4	
5	Connecting Plate	1	19154A
6	Screw, #10 x 1/2"	4	
7	Angle	1	19147B
8	Top Cover, 13"	1	20015B



Model 4007 Rack Mounting Kit

Parts List:

#### Description:

The Model 4007 is a rack mounting kit with overall dimensions, 7 inches high x 19 inches wide. Two top covers are provided for use with either 10 inch or 13 inch deep instruments.

#### Application:

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The Model 4007 converts the instrument from bench mounting to rack mounting. It is suitable for mounting two instruments in a standard 19-inch rack.

Item		Qty. Per	Keithley
No.	Description	Assembly	Part No.
1	Top Cover, 10"	2	20016B
4	Screw, #10 x 1/2"	8	
5	Connecting Plate	1	19154A
6	Screw, #10 x 1/2"	4	
7	Angle	2	14147B
8	Top Cover, 13"	2	200 <b>15</b> B
9	Zee Bracket	1	19167A
10	Plate	1	19700A

## SECTION 6. REPLACEABLE PARTS

6-1. REPLACEABLE PARTS LIST: This section contains a list of components used in this instrument for user reference. The Replaceable Parts List describes the individual parts giving Circuit Designation, Description, Suggested Manufacturer (Code Number), Manufacturer's Part Number, and the Keithley Part Number. Also included is a Figure Reference Number where applicable. The complete name and address of each Manufacturer is listed in the CODE-TO-NAME Listing following the parts list.

TABLE 6-1. Abbreviations and Symbols

A	ampere	F	farad	Ω	ohm
		Fig	Figure		
CbVar	Carbon Variable	-	5	р	pico (10 <sup>-12</sup> )
CerD	Ceramic Disc	GCb	Glass enclosed Carbon	PC	Printed Circuit
Cer Trimmer	Ceramic Trimmer			Poly	Polystyrene
Comp	Composition	k	kilo (10 <sup>3</sup> )	,	,,
			· · ·	Ref.	Reference
DCb	Deposited Carbon	μ	micro (10 <sup>-6</sup> )		
Desig.	Designation	i i		TCu	Tinner Copperwel
		м	Meg (10 <sup>6</sup> )		
EAL	Electrolytic, Aluminum	Mfr.	Manufacturer	v	volt
ETB	Electrolytic, tubular	MtF	Metal Film		
ETT	Electrolytic, tantalum	My	Mylar	W	watt
		, i	-	WW	Wirewound
		No.	Number	WWVar	Wirewound Variab

6-2. ELECTRICAL SCHEMATICS AND DIAGRAMS. Schematics and diagrams are included to describe the electrical circuits as discussed in Section 3.

6-3. HOW TO USE THE REPLACEABLE PARTS LIST. This Parts List is arranged such that the individual types of components are listed in alphabetical order. Main Chassis parts are listed followed by printed circuit boards and other subassemblies.

6-4. HOW TO ORDER PARTS.

a. Replaceable parts may be ordered through the

Sales Service Department, Keithley Instruments, Inc. or your nearest Keithley representative.

b. When ordering parts, include the following information.

- 1. Instrument Model Number.
- 2. Instrument Serial Number.
- 3. Part Description.
- 4. Schematic Circuit Designation.
- 5. Keithley Part Number.

c. All parts listed are maintained in Keithley Spare Parts Stock. Any part not listed can be made available upon request. Parts identified by the Keithley Manufacturing Code Number 80164 should be ordered directly from Keithley Instruments, Inc.

# MODEL 604 REPLACEABLE PARTS LIST (Refer to Schematic Diagram 22820E for circuit designations)

			CAF	ACITORS			
Circuit Desig.	Value	Rating	Туре	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
C101 C102 C103 C104 C105	10 μF 100 μF 100 μF 100 μF 100 μF	6 V 40 V 40 V 40 V 40 V	ETT EAl EAl EAL EAL	07436 73445 73445 73445 73445 73445	NG106A C437ÅR/G100 C437AR/G100 C437AR/G100 C437AR/G100	C182-10M C150-100M C150-100M C150-100M C150-100M	16 16 16 16 16
C106 C107 C108 C109 C110	100 μF 100 μF .001 μF 1.2 μF 100 μF	40 V 40 V 600 V 20 V 40 V	EAl EAl CerD ETT EAl	73445 73445 72982 05397 73445	C437AR/G100 C437AR/G100 ED001 K1R2J20K C437AR/G100	C150-100M C150-100M C22001M C80-1.2M C150-100M	16 16 16 16 16
Cl11 Cl12 Cl13 Cl14 Cl15	100 µF .001 µF 1.2 µF 7-25 pF 10 pF	40 V 600 V 20 V 200-300 V 600 V	EAl CerD ETT CerD	73445 72982 05397 72982 72982	C437AR/G100 ED001 K1R2J20K 538-037 ED-10	C150-100M C22001M C80-1.2M C175-7-25P C22-10P	16 16 16 16 16
C116 C117 C118 C119 C120	5 pF 22 pF Not Used 10 pF 33 pF	600 V 600 V 600 V 600 V	CerD CerD CerD CerD	72982 72982 72982 72982 72982	ED-5 ED-22 ED-10 ED-33	C22-5P C22-22P C22-10P C22-33P	16 14 14 14
C121 C122 C123 C124 C125	500 pF 1500 pF .005 µF .015 µF .05 µF	600 V 600 V 600 V 600 V 200 V	CerD CerD CerD CerD	72982 72982 72982 72982 72982 56289	ED~500 ED-1500 ED.005 ED.015 2PS-S50	C22-500P C22-1500P C22005M C22015M C17605M	15 15 15 15
C126 C127	.15 µF *10 pF	75 V 600 V	Poly Ceru	56289 72982	225P1549R752A3 2D-10	C16715M C22-10P	15

CONNECTORS AND TERMINALS

Circuit Desig.	Description	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
J101 J102	Receptacle, triaxial, +INPUT Receptacle, triaxial, -INPUT Plug, triaxial, mate of of J101	95712 95712	33050-nt-34 33050-nt-34	CS-181 CS-181	1
	and $J102$ (2 reg'd)	95712	30197-1	CS-141	
J103	Receptacle, microphone, AMPLIFIER				
	OUTPUT	02660	80-PC2F	CS-32	2
	Plug, microphone, mate of J103	02660	80-MC2M	CS-33	
J104	Receptacle, BNC, FROM -INPUT UNITY				
	GAIN OUTPUT (Mil. No. UG-1094/U)	02660	31-221	CS-15	2
J105	Receptacle, NC, FROM +INPUT UNITY				
	GAIN OUTPUT (Mil. No. UG-1094/U)	02660	31-221	CS-15	2
	Plug, BNC, mate of J104 and J105, 2				
	req'd (Mil. No. UG-88/U)	02660	31-002	CS-44	
J106	Card edge connector to AR101	09922	PSC44SS15-12	CS-175	
J107	Card edge connector to AR102	09922	PSC44SS15-12	CS-175	·
P101	Plug, coaxial, -Input of AR101	02660	FXR30775	CS-179	
	Hooded Shield for P101	02660	FXR2275	CS-180	
	Coaxial Receptacle, mate of P101	02660	FXR2175	CS-178	
P102	Plug, coaxial, +Input of AR101	02660	FXR30775	CS-179	
	Hooded hield for P102	02660	FXR2275	CS-180	
	Coaxial Receptacle, mate of P102	02660	FXR2175	CS-178	
P103	Plug, coaxial, -Input of AR102	02660	FXR30775	CS-179	
	Hooded Shield for P103	02660	FXR2275	CS-179 CS-180	
	Coaxial Receptacle, mate of P103	02660	FXR2175	CS-180 CS-178	
	countar moceptuoite, mate or rivs	02000	TVLTTA	C2=1/0	

\*Nominal Value, factory set.

### CONNECTORS AND TERMINALS (Cont'd)

Circuit Desig.	Description	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
P104	Plug, Coaxial, +Input of ARl02 Hooded Shield for Pl04 Coaxial Receptacle, mate of Pl04	02660 02660 02660	FXR30775 FXR2275 FXR2175	CS-179 CS-180 CS-178	 
	Binding Post, ground	08811	33-286	BP-15	

#### DIODES

Circuit			Mfg.	Keithley	Fig.
Desig.	Туре	Number	Code	Part No.	Ref.
D101	Silicon	1N645	01295	RF-14	16
D102	Silicon	1N645	01295	RF-14	16
D103	Silicon	1N645	01295	RF-14	16
D104	Silicon	1N645	01295	RF-14	16
D105	Silicon	1N645	01295	RF-14	16
DIUJ	01110011	11015	01293		+0
D106	Silicon	1N645	01295	RF-14	16
D107	Silicon	1N645	01295	RF-14	16
D108	Silicon	1N645	01295	RF-14	16
D109	Silicon	1N645	01295	RF-14	16
D110	Silicon	1N645	01295	RF-14	16
			••		
D111	Silicon	1N645	01295	RF-14	16
D112	Silicon	1N645	01295	RF-14	16
D113	Zener	1N715	12954	DZ-22	16
D114	Zener	1N715	12954	DZ-22	16
D115	Zener	1N715	12954	DZ-22	16
					_ •
D116	Zener	1N715	12954	DZ-22	16
D117	Zener	1N715	12954	DZ-22	16
D118	Zener	1N715	12954	D2-22	16
D119	Zener	1N715	12954	DZ-22	16
D120	Zener	1N715	12954	DZ-22	16
D121	Zener	1N702A	01295	DZ-33	16
D122	Zener	1N709	12954	DZ-21	16
D123	Silicon	1N645	01295	RF-14	īć
D124	Zener	1N709	12954	DZ-21	
D125	Silicon	1N645	01295	RF-14	16
		2110 10	0		20
D126	Zener	1N702A	01295	DZ-33	16
D127	Zener	1N935	04713	DZ - 7	16
D128	Zener	1N935	04713	DZ - 7	16
			0., 1.	55 ,	± 9

# INTEGRATED CIRCUITS AND OPERATIONAL AMPLIFIERS (For Parts List of AR101 and AR102, See Supplied Model 301 Manual)

Circuit Desig.	Description	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
AR101	Operational Amplifier to +INPUT Receptacle	80164	Model 301	Model 301	14
AR102	Operational Amplifier to -INPUT Receptacle	80164	Model 301	Model 301	14
QA101.	Integrated Circuit, Differential Amplifier	12040	LM301AH *	27225A	16
QA102	Integrated Circuit, Auxiliary Amplifier	12040	LM301AH *	27225A	16
QA103	Integrated Circuit, Output Amplifier	12040	LM301AH *	27225A	16

\*Selected for slew rate of 0.5V per Microsecond.

### MISCELLANEOUS PARTS

Circuit Desig.	Description	Mfg. Code	Keithley Part No.	Fig. Ref.
DS-101	Pilot Light (Mfg. No. 2100)	91802	PL-38	1
F101 (117V)	Fuse, Slow Blow, 1/8 A (Mfg. Type HDL)	71400	FU-20	2
F101 (234V)	Fuse, Slow Blow, l/l6 A (Mfq. Type HDL) Fuse Holder (Mfg. No. 342012)	75915	FH-3	
M101	Meter	80164	ME-83A	14
P105	Cord Set, 6 feet (Mfg. No. 4638-13) Cable Clamp (Mfg. No. SR-SP-1)	93656 28520	CO-5 CC-4	2
<b>T101</b>	Transformer	80164	TR-117	16

#### RESISTORS

Circuit Desig.	Value	Rating	Туре	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
R101 R102 R103 R104	**47 kΩ 1 kΩ 1 kΩ 1 kΩ	10%, 1/2 W 10%, 1/2 W 10%, 1/2 W	Comp Comp Comp	01121 01121 01121	EB EB EB	Rl-1K Rl-1K Rl-1K	17
R104 R105	$1 k\Omega$	10%, 1/2 W 10%, 1/2 W	Comp	01121	EB	RI-IK	17 17
R106 R107 R108 R109 R110	2.2 kΩ 2.2 kΩ 1 kΩ 15 Ω 1 kΩ	10%, 1/2 W 10%, 1/2 W 10%, 1/2 W 10%, 1/2 W 10%, 1/2 W	Comp Comp Comp Comp Comp	01121 01121 01121 01121 01121	EB EB EB EB	R1-2.2K R1-2.2K R1-1K R1-15 R1-1K	17 17 17 17 17 17
R111 R112 R113 R114 R115	5.6 kΩ 2 kΩ 4.7 kΩ 2.2 kΩ 2.2 kΩ	10%, 1/2 W 1/4 W 10%, 1/2 W 10%, 1/2 W 10%, 1/2 W	Comp Cb Comp Comp Comp	01121 37942 01121 01121 01121 01121	EB MTC-1.4 EB EB EB	R1-5.6K RP67-2K R1-4.7K R1-2.2K R1-2.2K	17 17 17 17 17
R116 R117 R118 R119 R120	1 kΩ 15 Ω 1 kΩ 4.7 kΩ 2 kΩ	10%, 1/2 W 10%, 1/2 W 10%, 1/2 W 10%, 1/2 W 10%, 1/2 W 1/4 W	Comp Comp Comp Comp Cb	01121 01121 01121 01121 37942	EB EB EB MTC-L4	R1-1K R1-15 R1-1K R1-4.7K RP67-2K	17 17 17 17 17
R121 R122 R123 R124 R125	5.6 kΩ 10 kΩ 4.02 kΩ 4.02 kΩ 4.02 kΩ	10%, 1/2 W ±5%, 2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W	Comp WWVar MtF MtF MtF	01121 12697 07716 07716 07716	EB Series 62 CEC CEC CEC	R1-5.6K RP70-10K R94-4.02K R94-4.02K R94-4.02K	14 15 15 15
R126 R127 R128 R129 R130	4.02 kΩ 4.02 kΩ 4.02 kΩ 4.02 kΩ 4.02 kΩ	1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W	MtF MtF MtF MtF MtF	07716 07716 07716 07716 07716 07716	CEC CEC CEC CEC CEC	R94-4.02K R94-4.02K R94-4.02K R94-4.02K R94-4.02K	15 15 15 15 15
R131 R132 R133 R134 R135	4.02 kΩ 15 kΩ 15 kΩ 15 kΩ 15 kΩ 15 kΩ	1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W	MtF MtF MtF MtF MtF	07716 07716 07716 07716 07716 07716	CEC CEC CEC CEC CEC	R94-4.02K R94-15K R94-15K R94-15K R94-15K	15 15 15 15 15
R136 R137 R138 R139 R140	15 kΩ 15 kΩ 15 kΩ 15 kΩ 15 kΩ	1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W	MtF MtF MtF MtF MtF	07716 07716 07716 07716 07716	CEC CEC CEC CEC CEC	R94-15K R94-15K R94-15K R94-15K R94-15K	15 15 15 15 15
R141 R142 R143 R144 R145	402 kΩ 402 kΩ 801 Ω 801 Ω 10 kΩ	1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 1%, 1/2 W 0.1%, 1/2 W	MtF MtF MtF MtF MtF	07716 07716 07716 07716 07716 07716	CEC CEC CEC CEC CEC T-1	R94-402K R94-402K R94-801 R94-801 R135-10K	17 17 17 17 17

\*\*Resistor R101 is an integral part of lamp DS101 and may be obtained by ordering this lamp from Keithley Instruments under the part number PL-38.

					-		
Circuit Desig.	Value	Rating	Туре	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
R146	10 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-10K	17
R147	99.5 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-99.5K	17
R148	$1 \ k\Omega$	0.1%, 1/4 W	Cb	37942	MTC-L4	RP67-1K	17
R149	$100 \ k\Omega$	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-100K	17
R150	3.45 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-3.45K	17
R151	100 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-100K	17
R152	10 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-10K	17
R153	<b>900</b> Ω	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-900	14
R154	<b>300</b> Ω	0.1%, 1/2 W	MtF	07716		R135-300	14
R155	l kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-1K	14
R156	3 kΩ	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-3K	14
R157	10 k $\Omega$	0.1%, 1/2 W	MtF	07716	CEC T-1	R135-10K	14
R158	2 kΩ	1/4 W	Cb	37942	MTC-L4	RP67-2K	17
R159	28.7 kΩ	1%, 1/8 W	MtF	07716	CEA	R88-28.7K	17
R160	9.76 kΩ	1%, 1/8 W	MtF	07716	CEA	R88-9.76K	17
R161	l kΩ	1/4 W	Cb	37942	MTC-L4	RP67-1K	17
R162	47_Ω	10%, 1/2 W	Comp	01121	EB		± /
R163	10 <sup>y</sup> Ω	20%, 1/2 W	Comp	75042	GBT	R1-47 R37-109	
R164	$479^{\Omega}$ 109 $\Omega$ 109 $\Omega$	20%, 1/2 W	Comp	75042	GBT	R37-10 <sup>9</sup>	
			-	-			

### RESISTORS (Cont'd)

### STRUCTURAL PARTS

Circuit Desig.	Description	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
	Top Cover Assembly	80164		20906B	
	Bottom Cover Assembly	80164	***	19298C	
	Foot, 4 req'd	72512	4004A	FE-5	
	Ball, one for each foot	80164		FE-6	

### SWITCHES AND CONTROLS

Circuit	Description	Mfg.	Keithley	Fig.
Desig.		Code	Part No.	Ref.
S101	Push Button Switch, +INPUT ZERO CHECK	80164	14376A	1
	Knob, +Input Zero Check Button	80164	14376A	
s102	Push Button Switch, -INPUT ZERO CHECK	80164	14376A	1
	Knob, -Input Zero Check Button	80164	14376A	1
S103	Slide Switch, GUARD-OFF	80164	SW-45	1
S104	Slide Switch, LINE VOLTAGE	80164	SW-151	2
s105	Rotary Switch, METER	80164	SW-165	L
	Knob Assembly, Meter Switch	80164	14838A	l
S106	Potentiometer, FINE ADJ. (R122) Knob, Fine Adj.	12697 80164	RP70-10K 16373A	1
S107 (A&B)  	Rotary Switch less components, MEDIUM and COARSE Rotary Switch with components, Medium and Coarse Knob Assembly, Medium Switch Knob Assembly, Coarse Switch	80164 80164 80164 80164	SW-267 22973b 16993A 16995A	1  1 1
S108	Rotary Switch less components, VOLTS	80164	SW-268	1
	Rotary Switch with components, Volts	80164	22980B	
	Knob Assembly, Volts Switch	80164	14838A	1
s109 	Rotary Switch less components, HIGH CUT HZ Rotary Switch with components, High Cut Hz Knob Assembly, High Cut Hz Switch	80164 80164 80164	SW-266 22974B 14838A	1

## TRANSISTORS

Circuit Desig.	Number	Mfg. Code	Keithley Part No.	Fig. Ref.
Q101 Q102 Q103 Q104 Q105	2N3565 2N3638 2N3565 2N3638 40317	07263 07263 07263 07263 07263 02734	TG-39 TG-33 TG-39 TG-33 TG-43	16 16 16 16 16
Q106 Q107 Q108 Q109 Q110	2N 3565 2N 3565 2N 3638 40319 2N 3638	07263 07263 07263 07263 02734 07263	TG-39 TG-39 TG-33 TG-50 TG-33	16 16 16 16 16

### MODEL 6041 REPLACEABLE PARTS LIST

(Refer to Schematic Diagram 23458C for circuit designations)

### CONNECTORS AND TERMINALS

Circuit Desig.	Description	Mfg. Code	Mfg. <u>Part No</u>	Keithley Part No.	Fig. Kei:
J101 	Receptacle, triaxial, +INPUT Plug, triaxial, mate of Jl01 Cap, +INPUT Receptacle	95712 95712 02660	33050-NT-34 30197-1 31-007	CS-181 CS-141 CAP-18	3
J102	Receptacle, triaxial, -INPUT Plug, triaxial, mate of J102 Cap, -INPUT Receptacle	95712 95712 02660	33050-NT-34 30197-1 31-007	CS-181 CS-141 CAP-18	3
J103	Binding Post, GND	08811	33-286	BP-15	3
P101	Connector, triaxial, TO 604 INPUT:+ Connector Assembly with cable	95712 80164	30197-1	CS-141	
P102	Connector, triaxial, TO 604 INPUT:- Connector Assembly with cable	95712 80164	30197-1	CS-141	3

## RESISTORS\*\*

Circuit Desig.	Value	Rating	Туре	Mfg. Code	Mfg. Part No.	Keithley Part No.	Fig. Ref.
R101	1011 N		~ ~1			······	
		+3-0%, 1/R W	GCb	63060	RX-1	R20-1011	
R102	$1010$ $\Omega$ $109$ $\Omega$	+3-0%, l/R W	GCb	63060	RX-1	R20-10-0	
R103	10° Ω	+3-0%, l/R W	GCb	63060	RX-1	R20-1010 R20-109	
R104	100MΩ	1%, 2W	DCb	91637	DC-2	R14-100M	
R105	10 MΩ	1%, 1/2W	DCb	91637	DCF-1/2	R12-10M	
R106	1 ΜΩ 1011 Ω 1010 Ω 109 Ω	1%, 1/2W	DCb	91637	DCF-1/2	R12-1M	
R107	10 <sup>⊥⊥</sup> Ω	+3-0%, 1/R W	GCb	63060	RX-1	R12-1M R20-1011	
R108	10 <sup>10</sup> Ω	+3-0%, 1/R W	GCb	63060	RX-1	R20-10 <sup>10</sup>	
R109	10 <sup>9</sup> Ω	+3-0%, 1/R W	GCb	63060	RX-1	$R_{20} - 10^9$	
R110	100MΩ		DCD				
NITO .	7.0.01488	1%, 2W	DCD	91637	DC-2	R14-100M	
R111	10 MΩ	18, 1/2W	DCb	91637	DCF-1/2	R12-10M	<u> </u>
R112	1 MΩ	1%, 1/2W	DCb	91637	DCF-1/2	R12-1M	~~

#### SWITCHES AND CONTROLS

Circuit Desig.	Description	Mfg. Code	Keithley Part No.	Fig. Ref
S101	Rotary Switch, +INPUT Knob Assembly, +Input Switch	80164 80164	Model 3011 16338A	3
s102	Rotary Switch, -INPUT Knob Assembly, -Input Switch	80164 80164	Model 3011 16338A	3

\*\* A complete set of resistor for the Model 6041 may be purchased from Keithley Instruments under the Model Number of 6033. The Model 6033 contains resistor pairs matched to within  $\pm 2$ %. It is advised that upon ordering resistors for the Model 6041 the user purchase matched pairs to assure accuracy.

TABLE 14. Code List of Suggested Manufacturers. (Based on Federal Supply Code for Manufacturers, Cataloging Handbook H4-1).

- 01121 Allen-Bradley Corp. 1201 South 2nd Street Milwaukee, Wis. 53204
- 01295 Texas Instruments, Inc. Semiconductor-Components Division 13500 North Central Expressway Dallas, Texas 75231
- 02660 Amphenol Corp. 2801 South 25th Avenue Broadview, Ill. 60153
- 02734 Radio Corp. of America Defense Electronic Products Camden, N. J.
- 04713 Motorola Semiconductor Products Inc. 5005 East McDowell Road Phoenix, Ariz. 85008
- 05397 Union Carbide Corp. Linde Div. Kemet Dept. 11901 Madison Cleveland, Ohio 44107
- 07263 Fairchild Camera & Instru. Corp. Semiconductor Division 313 Frontage Road Mountain View, Cal.
- 07716 IRC, Inc. 2850 Mt. Pleasant Burlington, Iowa 52601
- 08811 GL Electronics Division of GL Industries Inc. 300 Harvard Avenue Westville, N. J. 08093
- 09922 Burndy Corp. Richards Avenue Norwalk, Conn. 06852
- 12040 National Semiconductor Corp. Commerce Drive Post Office Box 443 Danbury, Conn. 06813
- 12697 Clarostat Mfg. Co., Inc. Lower Washington Street Dover, N. H. 03820
- 12954 Dickson Electronics Corp. 302 S. Wells Fargo Avenue Scottsdale, Arizona

- 28520 Heyman Mfg. Co. 147 N. Michigan Avenue Kenilworth, N. J.
- 37942 Mallory, P. R. and Co., Inc. 3029 E. Washington Street Indianapolis, Ind. 46206
- 56289 Sprague Electric Co. North Adams, Mass.
- 63060 Victoreen Instruments Co. 5806 Hough Avenue Cleveland, Ohio 44103
- 71400 Bussmann Mfg. Div. of McGraw-Edison Co. 2538 W. University Street St. Louis, Mo.
- 72982 Erie Technological Products, Inc. 644 W. 12th Street Erie, PA. 16512
- 73445 Amperex Electronic Co., Div. of North American Philips Co., Inc. Hicksville, N. Y.
- 75042 IRC, Inc. 401 North Broad Street Philadelphia, PA. 19108
- 75915 Littlefuse, Inc. 800 E. Northwest Hwy. Des Plaines, Ill. 60016
- 80164 Keithley Instruments, Inc. 28775 Aurora Road Cleveland, Ohio 44139
- 91802 Industrial Devices Inc. 982 River Road Edgewater, N. J. 07020
- 93656 Electric Cord Co. 1275 Bloomfield Avenue Caldwell, N. J.
- 95712 Dage Electric Co., Inc. Hurricane Road Franklin, Indiana

### Mechanical Parts List

	Description	Quantity Per Assembly	Keithley Part No.
1)	Chassis	1	22636B
11)	Front Panel	1	22306C
	Top Cover Assembly		20 <b>90</b> 6B
	12) Cover, Sheet Metal	1	24561D
	13) Screws	4	
	Handle Assembly		
	14) Handle	1	НН-18
	15) Screws #6-32 x 3/8" R.H. slotted	2	
	Bottom Cover Assembly		19298C
	2) Cover	I	17149C
	3) Fastener	2	FA-54
	Feet Assembly		
	4) Feet	4	FE-5
	5) Ball	4	FE-6
	6) Screws #8-32 x 3/8" Phillips, Pan Head	4	·
	Tilt Bail Assembly		
	7) Bail	1	17147B
	8) Right Assembly	l	19206B
	9) Left Assembly	1	19205B
	10) Screws #6-32 x 1/4" Phillips, Pan Head	2	

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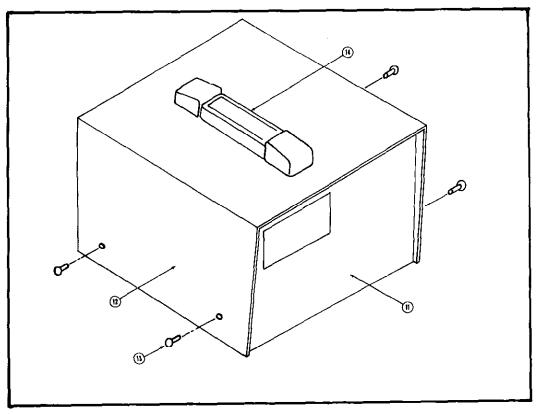


FIGURE 18. Top Cover Assembly.

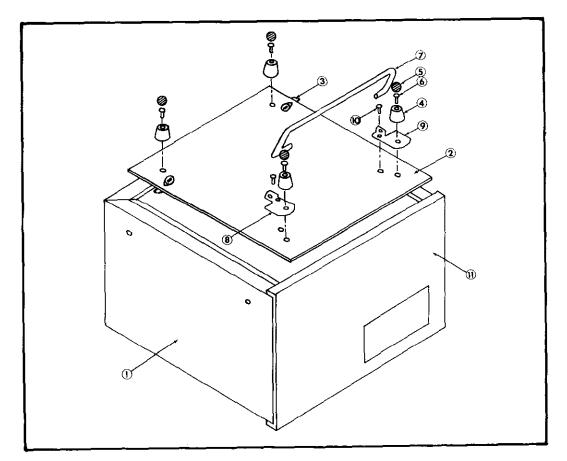


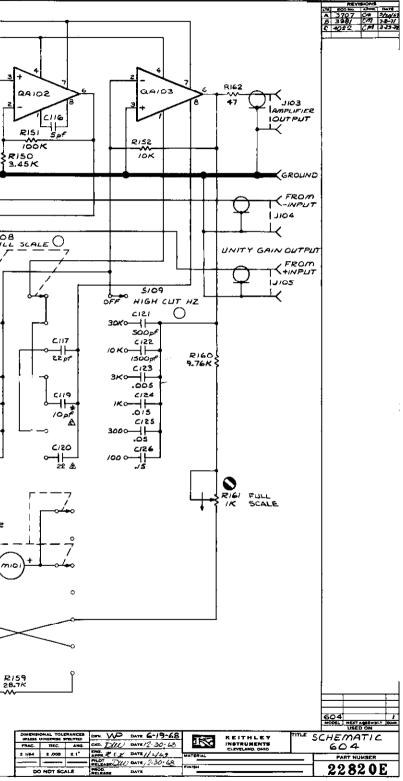
FIGURE 19. Bottom Cover Assembly.

01 R149 100K J 106 🛦 JIDI PIOI 511 +INPUT  $\sim$ R145 10K -∛≪ ⇒ SHIELD K ARIOI <u>\*</u>(( 2164 ୷≫ QAIDI R146 10K SIGI  $\square$ PIOZ GROU <u>-′°</u>≪ 5103 GUARD C101 R143 \$ ŧ <u>•</u>(( 124 SUPPRESS-SIOTB COARSE RIAI FINE 5107A MEDIUM DID3 0/1 JUD7 JIOZ -TURNI) 0.10 VOLTS FULL SCALE RIOZ ≫ \$R124 \$4,02K \$ R133 15K K ARIO2 SHIELDY ้กมะ D127 C102 + ≫ R163 SIOZ A DIIA L R125 S4.02K ₹ R134 \$ 15K I GROUND C103 + D115 余 .001 R126 2 R135 R/53 900 .003 2127 | 4.02Ki 2 R136 POWER OFF aioz D105 D107 .01 0103 ŽRIZB ( \$4.02K R137 <u>ليماري</u> R/54 300 LINE VOLTAGE OFF Ð DIOG R104 / DI28 ØГ .03 🖌 DIO8 اەندە ( RIZ9 \$4.02K \$ R138 ₹155 /K •-₩-5104 דעסי + 100 T 8110 ./ 0 234V --\_ 8130 \$ R139 R/56 3K L DII9 / ZERO CENTER R101 47K 5) R105 .3 סבים 🖊 FIOI \$ R140 \$ 15K R157 10K a104 1 R144 801 R142 0109 ווום R109 15 ( Pios aios ₹<u>₹</u>/06 2.2K الم . 0110 SUD Q107 SIO**S** METER 8 R II I 5 G K  $\overline{}$ R107 2.2K 1 cioa T 2001 \$ R110 K RII2+15V 2K ADJ. DI23 (106 + NOTES: 上+ C109 丁 1.2 R108 1K ALL RESISTANCE & CAPACITANCE SHALL BE DESIGNATED IN OHMS & MICROFARADS UNLESS OTHERWISE NOTED. тюі 🖻 Ι. POWER 0106/ OFF 2. C FRONT PANEL CONTROL. OFFO 5R113 4.7K DIZZ S INTERNAL SCREWDRIVER ADJ. З.  $\odot$ REAR PANEL CONTROL. Ц. Z #119 24.7K L DIZ4 1000 OHM 5. K C/// +1 6. pf PICOFARAD RIGALST REFERENCE DESIGNATIONS RIG4 CI27 DI28 Q110 J107 SIOS M101 P105 DSI01 F101 T101 GA103 KIIG IK LCII2 T-001 7. 🛉 CLOCKWISE ROTATION. ZERO CENTER 1+c113 T 1.2 RIIB RISB ZK R114 2.2K INDICATES NOMINAL VALUE DETERMINED DURING FINAL CALIBRATION 8. \* \$ R121 \$ 5.6K R115 R117 109 REFERENCE DESIGNATIONS **~** DIZG сна

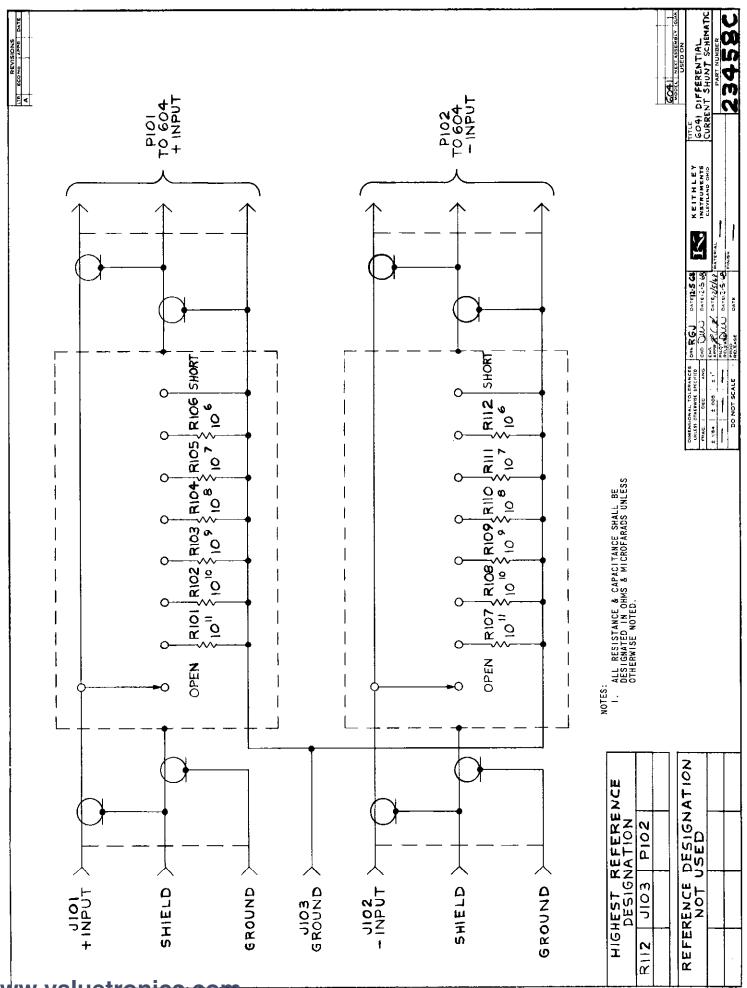
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CN604-2 Applies to Manual Printed January, 1977.



## KEITHLEY INSTRUMENTS, INC.

## INSTRUCTION MANUAL CHANGE NOTICE MODEL 604 ELECTROMETER AMPLIFIER

INTRODUCTION: Since Keithley Instruments is continually improving product performance and reliability, it is often necessary to make changes to Instruction Manuals to reflect these improvements. Also, errors in Instruction Manuals occasionally occur that require changes. Sometimes, due to printing lead time and shipping requirements, we can't get these changes immediately into printed Manuals. The following new change information is supplied as a supplement to this Manual in order to provide the user with the latest improvements and corrections in the shortest possible time. Many users will transfer this change information directly to a Manual to minimize user error. All changes or additions are indicated in *italics*.

CHANGES:

CI14 1.9pF-15.7pF 200-300 V 13050 C-284-1.9P-15.7P 16

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	KEITHLEY INSTRUMENTS, INC. 28775 AURORA ROAD CLEVELAND, OHIO 44139 SERVICE FORM
NAME	NO SERIAL NO P.O. NO DATE R PHONE
	SS CITY STATE ZIP
1.	Describe problem and symptoms using quantitative data whenever possible (enclose readings, chart recordings, etc.)
2.	(Attach additional sheets as necessary). Show a block diagram of your measurement system including all instruments connected (whether power is turned on or not). Also describe signal source.
3.	List the positions of <u>all</u> controls and switches on both front and rear panels of the instrument.
4.	Describe input signal source levels, frequencies, etc.
5.	List and describe all cables used in the experiment (length, shielding, etc.).
6.	List and describe all other equipment used in the experiment. Give control settings for each.
7.	Environment: Where is the measurement being performed? (Factory, controlled laboratory, out-of-doors, etc.) What power line voltage is used?Variation?Frequency? Ambient temperature?°F. Variation?°F. Rel. Humidity?
8.	OtherAdditional Information. (If special modifications have been made by the user, please describe below.)